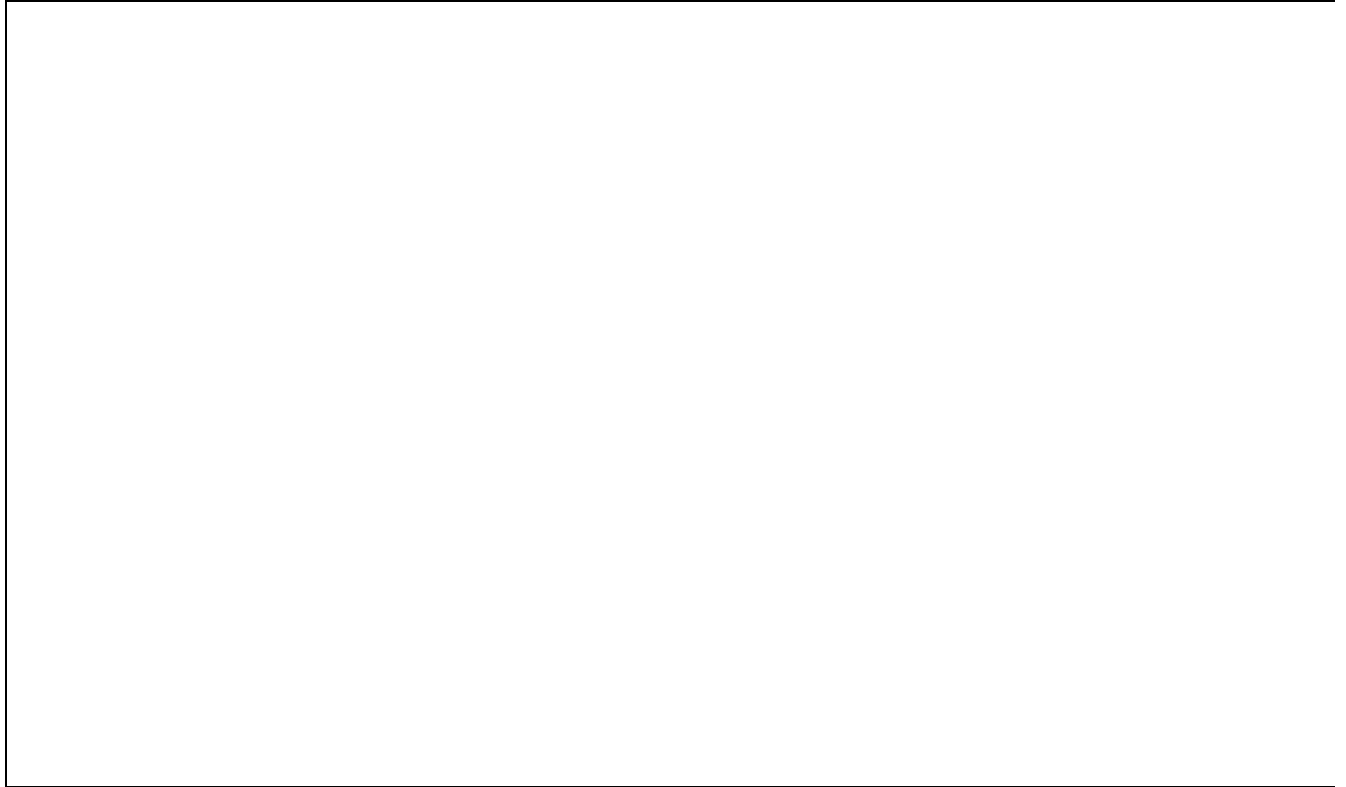


SIEMENS



ICs for Communications

Mixer/Amplifier

PMB 2333 Version 1.2

Preliminary Data Sheet 09.97

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| | | |
|-------------------------------|-----------------------------|--|
| PMB 2333 | | |
| Revision History: | | Current Version: 09.97 |
| Previous Version: | | 06.96 |
| Page (in 06.96 Version) | Page (in new Version) | Subjects (major changes since last revision) |
| 10 | 10 | Supply Voltage -> 5.0V max. |
| 10 | 10 | Input Voltage $V_{LO/X}$ -> 5.0V max. |
| 10 | 10 | Input Voltage V_{AI} -> $V_{AO}+0.3V$ max. |
| | 10 | Input Voltage V_{AI} AC Peak -> -2V min. |
| 10 | 10 | Input Voltage V_{GC} -> -0.3V min. / $V_S+0.3$ max. |
| 10 | 10 | Input Voltage V_{STB} -> 5.0V max. |
| 10 | 10 | Open Collector Output Voltage $V_{MO/X}$ ->1.7V min. / 5.0V max. |
| | 10 | Amplifier Current (Base, Peak) I_{AI} -> 6mA |
| | 10 | Amplifier Power dissipation P_{AMPtot} -> 105mW |
| 10 | 11 | Thermal Resistance R_{thJA} -> 213K/W |
| | 11 | Thermal Resistance R_{thSO} -> 160K/W |
| | 11 | ESD Integrity |
| 25 | 26 | Amplifier $f=0.9GHz$ -> Γ_{OPT} |
| 26 | 27 | Amplifier $f=1.8GHz$ -> Γ_{OPT} |
| 32 | 33 | Diagram 5 - Identical Values - New Printout |
| | | Update Of / Additional Application Information |
| | | Correction Of Printing Mistakes |

| | | |
|-------------------------------|-------------------------------|---|
| PMB 2333 | | |
| Revision History: | | Current Version: 06.96 |
| Previous Version: | | 11.95 |
| Page (in 11.95 Version) | Page (in 06.96 Version) | Subjects (major changes since last revision) |
| | | Update of RF/S-parameters because of cavity change, correction of printing mistakes, update of application circuits |
| | | |

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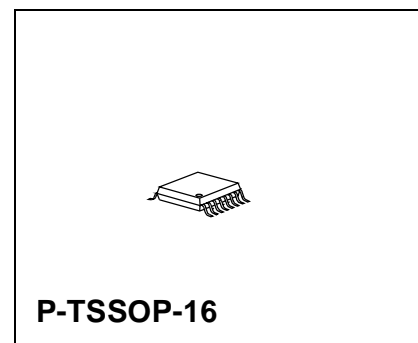
1 Overview

1.1 Functional Description, Benefits

- New B6HF bipolar technology, 25GHz f_T
- Small outline P-TSSOP 16 package
- Reduced external components
- Frequency range up to 3.0GHz
- Amplifier may be used as LNA or Driver
- LNA mode
 - 1.7dB typ. noise figure at 1.8GHz
 - 5mA typ. current consumption
- Driver mode
 - +12dBm output at 1dB compression
 - 20mA current consumption
- Gilbert cell mixer with high gain
- 2.7 - 4.5V voltage supply
- -40°C to +85°C operational temperature range
- Standby function
- High isolation values for amplifier and mixer
- Good crosstalk performance

1.2 Applications

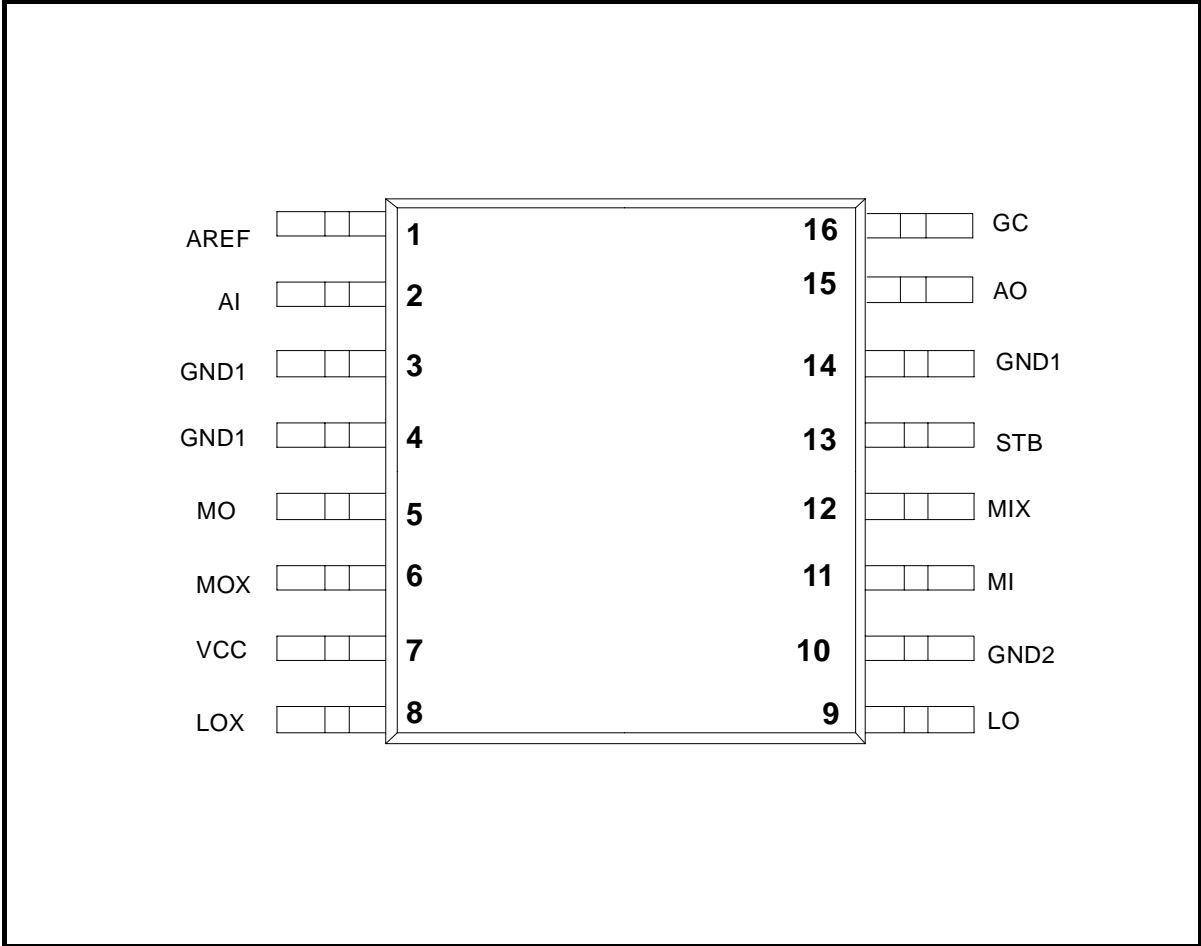
- Cellular radio systems
- Cordless telephone systems
- WLAN-Systems



| Type | Version | Ordering Code | Package |
|----------|---------|---------------|------------|
| PMB 2333 | V1.2 | Q67006-A6128 | P-TSSOP-16 |

1.3 Pin Configuration

(top view)

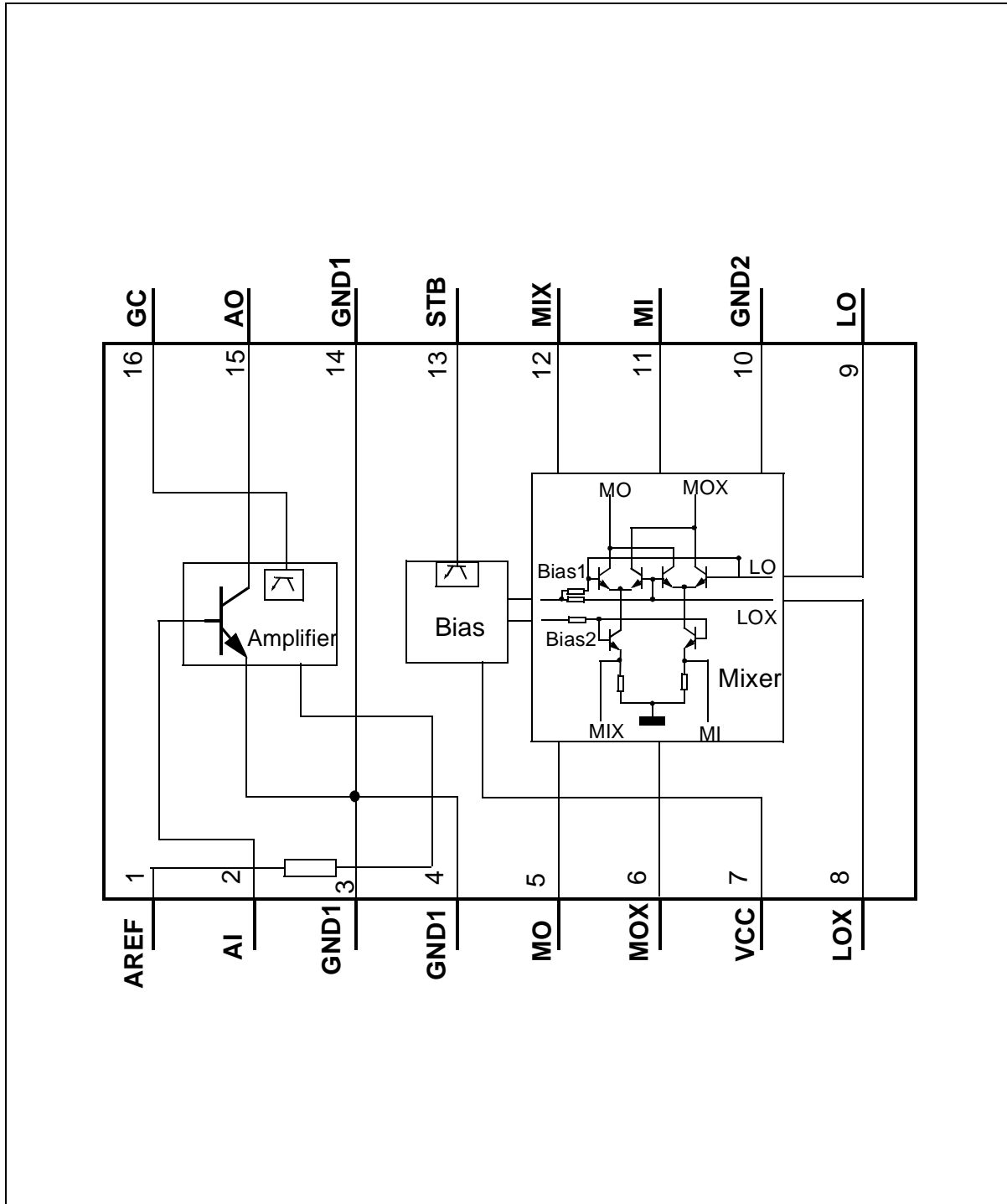


P-TSSOP16

1.4 Pin Definitions and Functions

| Pin No. | Symbol | Function |
|---------|--------|--|
| 1 | AREF | Amplifier bias supply for signal input |
| 2 | AI | Amplifier signal base input |
| 3 | GND1 | Amplifier ground |
| 4 | GND1 | Amplifier ground |
| 5 | MO | Mixer signal open collector output |
| 6 | MOX | Mixer signal open collector output |
| 7 | VCC | Voltage supply total circuit |
| 8 | LOX | Mixer local oscillator signal base input, inverted |
| 9 | LO | Mixer local oscillator signal base input, not inverted |
| 10 | GND2 | Mixer ground |
| 11 | MI | Mixer signal emitter input, not inverted |
| 12 | MIX | Mixer signal emitter input, inverted |
| 13 | STB | Standby mixer and bandgap |
| 14 | GND1 | Amplifier ground |
| 15 | AO | Amplifier signal open collector output |
| 16 | GC | Amplifier gain control |

1.5 Functional Block Diagram



1.6 Circuit Description

MIXER

The mixer used in this design is a general purpose up-/down conversion gilbert cell mixer. Via the pins MI/MIX the RF enters the IC. Using an external supplied local oscillator at LO/LOX a converted output signal is created at the open collector output pins MO/MOX. The open collector pins need to be connected to an external voltage supply. The RF connection to the mixer inputs can be single ended or balanced, capacitive or inductive. To improve the mixer performance external resistors at MI/MIX make it possible to adjust the mixer current. Voltage supply for the mixer has to be connected to the pin VCC and to GND2.

AMPLIFIER

The amplifier may be used as a low noise amplifier LNA or as a driver. At pin AI the RF signal enters the IC, at the open collector output AO, which need to be connected to supply voltage, the amplified signal is external available for further use. Matching networks at in-/and output can be used for improving the gain and the noise performance. To reduce the series feedback of the emitter line the amplifier is connected to ground via three GND1 pins. At AREF a internal supplied reference voltage is available for the DC biasing of AI. This dc output should be implemented in an input matching network. The voltage supply for the amplifier is also VCC. The dc-level at the pin GC allows to adjust the amplifier current.

Lower current is recommended for using the amplifier as an LNA, high current for using it as a driver.

COMMON

Differential signals and symmetrical circuits are used throughout the mixer part of the IC. An internal bias driver generates supply voltage and temperature compensated reference voltages. The STB pin allows the mixer and bandgap part of the IC to be switched in a low power mode.

All pins with the exception of GND1,2 and AI/AO are ESD protected.

2 Electrical Characteristics

2.1 Absolute Maximum Ratings

The maximum ratings may not be exceeded under any circumstances, not even momentarily and individually, as permanent damage to the IC will result.

Ambient temperature $T_{amb} = -40^{\circ}\text{C} \dots +85^{\circ}\text{C}$

| # | Parameter | Symbol | Limit Values | | Units | Remarks |
|----|--------------------------------|--------------|--------------|-------------------------|--------------------|-----------------------------|
| | | | Min | Max | | |
| 1 | Supply Voltage | V_S | -0.3 | 5.0 | V | |
| 2a | Input Voltage | $V_{MI/MIX}$ | -0.3 | 1.9 | V | $V_S = 0$ |
| 2b | Input Voltage | $V_{LO/LOX}$ | 0.6 | $V_S+0.3$ 5.0max. | V V | |
| 2c | Input Voltage | V_{AI} | -0.3 | $V_{AO}+0.3$ 3.5max. | V V | |
| 2d | Input Voltage (AC Peak) | V_{AI} | -2 | | V | Freq.>1MHz $I_{AI} < nA$ |
| 2e | Input Voltage | V_{GC} | -0.3 | $V_S+0.3$ 2.7max. | V V | |
| 2f | Input Voltage | V_{STB} | -0.3 | $V_S+0.3$ 5.0max. | V V | |
| 3a | Output Voltage | V_{AREF} | -0.3 | 2.0 | V | |
| 3b | Open Collector Output Voltage | $V_{MO/MOX}$ | 1.7 | $V_S+0.3$ 5.0max. | V V | |
| 3c | Open Collector Output Voltage | V_{AO} | -0.3 | 3.5 | V | Base open |
| 3d | Open Collector Output Voltage | V_{AO} | -0.3 | 7.0 | V | $R_B < 50k\Omega$ |
| 4a | Amplifier Current (Collector) | I_{AO} | | 30 | mA | |
| 4b | Amplifier Current (Base, Peak) | I_{AI} | | 6.0 | mA | DC and AC |
| 4c | Amplifier Power Dissipation | P_{AMPtot} | | 105 | mW | |
| 5 | Differential Input Voltage | V_{DIFF} | | 2.0 | V_{PP} | |
| 6 | Junction Temperature | T_j | | 125 | $^{\circ}\text{C}$ | |

Absolute Maximum Ratings

The maximum ratings may not be exceeded under any circumstances, not even momentarily and individually, as permanent damage to the IC will result.

Ambient temperature $T_{amb} = -40^{\circ}\text{C} \dots +85^{\circ}\text{C}$

| # | Parameter | Symbol | Limit Values | | Units | Remarks |
|----|---|------------|--------------|-----|--------------------|---------|
| | | | Min | Max | | |
| 7 | Storage Temperature | T_S | -40 | 125 | $^{\circ}\text{C}$ | |
| 8a | Thermal Resistance | R_{thJA} | | 213 | K/W | 1) |
| 8b | Thermal Resistance | R_{thSO} | | 160 | K/W | 2) |
| 9 | ESD integrity, all pins without AI, AO and GND1/2 | V_{ESD} | -500 | 500 | V | 3) |

1) Attention: **Do not exceed the max. junction temperature**

2) Junction to soldering point, simulated with FEM

3) HBM according MIL STD 883D, method 3015.7, and EOS/ESD assn. standard S5.1-1993

2.2 Operational Range

Within the operational range the IC operates as described in the circuit description.
The AC/DC characteristic limits are not guaranteed.

Supply voltage $V_{VCC} = 2.7V...4.5V$, Ambient temperature $T_{amb} = -40^{\circ}C...85^{\circ}C$

| # | Parameter | Symbol | Limit Values | | Units | Remarks |
|---|----------------------------|-------------|--------------|-------|-------|-----------|
| | | | Min | Max | | |
| 1 | AI Input Frequency | f_{AI} | | 3000 | MHz | |
| 2 | MI/X Input Frequency | f_{MI} | | 3000 | MHz | |
| 3 | LO/X Input Frequency | f_{LO} | | 3000 | MHz | |
| 4 | IF Intermediate Frequency | f_{IF} | | 3000 | MHz | |
| 5 | Standby Voltage On | STB_{ON} | 2.0 | V_S | V | |
| 6 | Standby Voltage Off | STB_{OFF} | 0 | 0.5 | V | |
| 7 | Gain Control Voltage, High | GC_{High} | 2.0 | 2.7 | V | Diagram 5 |
| 8 | Gain Control Voltage, Low | GC_{Low} | 0 | 0.6 | V | Diagram 5 |

Note: Power levels refer to 50 Ohms impedance

2.3 AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V...4.5V$, Ambient temperature $T_{amb} = +25^{\circ}C$

| # | Parameter | Symbol | Limit Values | | | Units | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|-------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

Supply Current

| | | | | | | | | |
|----|--------------------------|----------------|--|------|--|---------|---|----------------------|
| 1a | Supply current, total IC | $I_{5,6,7,15}$ | | 23.6 | | mA | STB ON, no external resistors at MI/MIX* | 1, $I_{AO}=20$ mA |
| 1b | Supply current, total IC | $I_{5,6,7,15}$ | | 1.6 | | mA | STB ON, no external resistors at MI/MIX* | 1, $I_{AO}=0$ mA |
| 2 | Supply current, total IC | $I_{5,6,7,15}$ | | <20 | | μA | STB OFF, GC=0V | 1 |

* Minimum value for external resistors at MI/MIX: $R1=R2=330\Omega$

AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V$ to $4.5V$, Ambient temperature $T_{amb} = +25^{\circ}$

| # | Parameter | Symbol | Limit Values | | | Unit | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

AMPLIFIER-Driver, Signal Input AI, $I_{AO}=20mA$, $V_{AO}=3.3V$, $f=2.5GHz$

| | | | | | | | | |
|---|-----------------------------------|--------------|--|------|--|-----|------------|---|
| 3 | Input impedance, vs. freq. | S_{11} | | | | | Diagram 3a | 3 |
| 4 | Max. input level, 1db compression | P_{AI} | | 0.0 | | dbm | f=2.5GHz | 1 |
| 5 | Input intercept, third order | $IICP_{DAI}$ | | 10.0 | | dbm | f=2.5GHz | 1 |

AMPLIFIER-Driver, Signal Output AO, $I_{AO}=20mA$, $V_{AO}=3.3V$, $f=2.5GHz$

| | | | | | | | | |
|---|----------------------------|-------------|--|-------|--|----|------------|---|
| 6 | Output current | I_{AO} | | 20.0 | | mA | | 1 |
| 7 | Output impedance vs. freq. | S_{22} | | | | | Diagram 3a | 3 |
| 8 | Power gain | S_{21Amp} | | +12.5 | | db | f=2.5GHz | 1 |

AMPLIFIER-Driver, Signal Input AI, $I_{AO}=0mA$, $V_{AO}=3.3V$, $f=2.5GHz$

| | | | | | | | | |
|----|------------------------------|-------------|--|------|--|-----|------------|---|
| 9 | Input impedance, vs. freq. | S_{11} | | | | | Diagram 3c | 3 |
| 10 | Max. input level, 1db change | P_{AI} | | 0.0 | | dbm | f=2.5GHz | 1 |
| 11 | Input intercept, third order | $IICP_{AI}$ | | 20.0 | | dbm | f=2.5GHz | 1 |

AMPLIFIER-Driver, Signal Output AO, $I_{AO}=0mA$, $V_{AO}=3.3V$, $f=2.5GHz$

| | | | | | | | | |
|----|----------------------------|-------------|--|-------|--|----|------------|---|
| 12 | Output current | I_{AO} | | 0 | | mA | | 1 |
| 13 | Output impedance vs. freq. | S_{22} | | | | | Diagram 3c | 3 |
| 14 | Power gain | S_{21Amp} | | < -13 | | db | f=2.5GHz | 1 |

All amplifier measurements have be done with Siemens RT5880 Duroid (Teflon) Boards

AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V$ to $4.5V$, Ambient temperature $T_{amb} = +25^{\circ}$

| # | Parameter | Symbol | Limit Values | | | Unit | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

AMPLIFIER-LNA, Signal Input AI, $I_{AO}=5mA$, $V_{AO}=3.3V$, $f=1.8GHz$

| | | | | | | | | |
|----|-----------------------------------|-------------|--|-------|--|-----|------------|---|
| 15 | Input impedance, vs. freq. | S_{11} | | | | | Diagram 3b | 3 |
| 16 | Max. input level, 1dB compression | P_{AI} | | -12.0 | | dBm | f=1.8GHz | 1 |
| 17 | Input intercept, third order | $IICP_{AI}$ | | 1.0 | | dBm | f=1.8GHz | 1 |
| 18 | Noise figure | F_{AI} | | 1.7 | | dB | f=1.8GHz | 1 |

AMPLIFIER-LNA, Signal Output AO, $I_{AO}=5mA$, $V_{AO}=3.3V$, $f=1.8GHz$

| | | | | | | | | |
|----|----------------------------|-------------|--|------|--|----|------------|---|
| 19 | Output current | I_{AO} | | 5.0 | | mA | | 1 |
| 20 | Output impedance vs. freq. | S_{22} | | | | | Diagram 3b | 3 |
| 21 | Power gain | S_{21LNA} | | 12.0 | | dB | f=1.8GHz | 1 |

AMPLIFIER-LNA, Signal Input AI, $I_{AO}=0mA$, $V_{AO}=3.3V$, $f=1.8GHz$

| | | | | | | | | |
|----|------------------------------|-------------|--|------|--|-----|------------|---|
| 22 | Input impedance, vs. freq. | S_{11} | | | | | Diagram 3c | 3 |
| 23 | Max. input level, 1db change | P_{AI} | | 0.0 | | dbm | f=1.8GHz | 1 |
| 23 | Input intercept, third order | $IICP_{AI}$ | | 20.0 | | dbm | f=1.8GHz | 1 |

AMPLIFIER-LNA, Signal Output AO, $I_{AO}=0mA$, $V_{AO}=3.3V$, $f=1.8GHz$

| | | | | | | | | |
|----|----------------------------|-----------|--|-------|--|----|------------|---|
| 24 | Output current | I_{AO} | | 0 | | mA | | 1 |
| 25 | Output impedance vs. freq. | S_{22} | | | | | Diagram 3c | 3 |
| 26 | Power gain | S_{21A} | | < -17 | | db | f=1.8GHz | 1 |

All LNA measurements have be done with Siemens RT5880 Duroid (Teflon) Boards

AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V$ to $4.5V$, Ambient temperature $T_{amb} = +25^{\circ}$

| # | Parameter | Symbol | Limit Values | | | Unit | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

MIXER, Signal Input MI/MIX, Upconversion, R1,2=330hm

| | | | | | | | | |
|----|------------------------------------|--------------|--|----|--|-----|------------------|----|
| 27 | Input impedance vs .freq. | Z_{MI} | | | | | Diagram 4a | 4 |
| 28 | Max. input level, 1 db compression | P_{MI} | | -7 | | dbm | $f_{MI}=0.66GHz$ | 1* |
| 29 | Input intercept point | $IICP3_{MI}$ | | 6 | | dbm | $f_{MI}=0.66GHz$ | 1* |

MIXER, Local Oscillator Input LO/LOX, Upconversion, R1,2=330hm

| | | | | | | | | |
|----|--------------------------|----------|--|---|--|-----|-----------------|----|
| 30 | Input impedance vs freq. | Z_{LO} | | | | | Diagram 4c | 4 |
| 31 | Input level | P_{LO} | | 0 | | dbm | $f_{LO}=2.0GHz$ | 1* |

MIXER, Signal Output MO/MOX, $f_{out} = 2.66GHz$, Upconversion, R1,2=330hm

| | | | | | | | | |
|----|--------------------|--------------|--|------|--|-----|-------------------------------|----|
| 32 | Output current | $I_{MO/X}$ | | 10.4 | | mA | with ext. resistors at MI/MIX | 1* |
| 33 | Output resistance | R_{MODiff} | | 600 | | Ohm | $f_{MO}=2.66GHz$ | |
| 34 | Output capacitance | C_{MODiff} | | 0.57 | | pF | $f_{MO}=2.66GHz$ | |
| 35 | Power gain | P_{MI} | | 8 | | db | $f_{MO}=2.66GHz$ | 1* |

MIXER, Isolation Between In-/Output, $f_{out} = 2.66GHz$, Upconversion, R1,2=330hm

| | | | | | | | | |
|----|----------|-------------|--|----|--|----|--|----|
| 37 | LO to MO | A_{LO-MO} | | 30 | | db | | 1* |
| 38 | LO to MI | A_{LO-MI} | | 35 | | db | | 1* |
| 39 | MO to MI | A_{MO-MI} | | 40 | | db | | 1* |
| 40 | MO to LO | A_{MO-LO} | | 45 | | db | | 1* |

* MI/MO Input/Output including matching network

AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V$ to $4.5V$, Ambient temperature $T_{amb} = +25^{\circ}$

| # | Parameter | Symbol | Limit Values | | | Unit | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

MIXER, Signal Input MI/MIX, Downconversion, R1,2=180Ohm

| | | | | | | | | |
|----|---------------------------------------|-----------------|--|-----|--|-----|------------|----|
| 41 | Input impedance vs .freq. | Z_{MI} | | | | | Diagram 4b | 4 |
| 42 | Max. input level, 1 db compression | P_{MI} | | -15 | | dBm | f=0.9GHz | 2a |
| 43 | at MO/MOX, IF=45MHz | P_{MI} | | -14 | | dBm | f=1.8GHz | 2a |
| 44 | | P_{MI} | | -9 | | dBm | f=2.5GHz | 2a |
| 45 | Input intercept point, | $IICP3_{MI}$ | | 0 | | dBm | f=0.9GHz | 2a |
| 46 | $\Delta f=800kHz$, IF= 45MHz | $IICP3_{MI}$ | | -1 | | dBm | f=1.8GHz | 2a |
| 47 | | $IICP3_{MI}$ | | +5 | | dBm | f=2.5GHz | 2a |
| 48 | Blocking level, | $P_{in,unwan.}$ | | -16 | | dBm | f=0.9GHz | 2a |
| 49 | $\Delta f=800kHz$, IF=45MHz | $P_{BL,unwan.}$ | | -16 | | dBm | f=1.8GHz | 2a |
| 50 | $P_{in, wanted} = -20dBm$ | $P_{BL,unwan.}$ | | -10 | | dBm | f=2.5GHz | 2a |
| 51 | Noise figure, ssb | F_{MI} | | 9 | | dB | f=0.9GHz | * |
| 52 | ($NF_{ssb} \approx NF_{dsb} + 3dB$) | F_{MI} | | 11 | | dB | f=1.8GHz | * |
| 53 | IF=45MHz | F_{MI} | | 14 | | dB | f=2.5GHz | * |

MIXER, Local Oscillator Input LO/LOX, Downconversion, R1,2=180Ohm

| | | | | | | | | |
|----|--------------------------|----------|--|----|--|-----|------------|--------|
| 54 | Input impedance vs freq. | Z_{LO} | | | | | Diagram 4d | 4 |
| 55 | Input level | P_{LO} | | -3 | | dBm | f=0.9GHz | 2a, ** |
| 56 | | P_{LO} | | -3 | | dBm | f=1.8GHz | 2a, ** |
| 57 | | P_{LO} | | -3 | | dBm | f=2.5GHz | 2a, ** |

* matching network used

** referenced for specified mixer performance

AC/DC Characteristics

AC/DC characteristics involve the spread of values guaranteed within the specified supply voltage and ambient temperature range. Typical characteristics are the median of the production.

Supply voltage $V_{VCC} = 2.7V$ to $4.5V$, Ambient temperature $T_{amb} = +25^{\circ}$

| # | Parameter | Symbol | Limit Values | | | Unit | Test Conditions | Test Circuit |
|---|-----------|--------|--------------|-----|-----|------|-----------------|--------------|
| | | | Min | Typ | Max | | | |

MIXER, Signal Output MO/MOX, Downconversion, R1,2=180Ohm

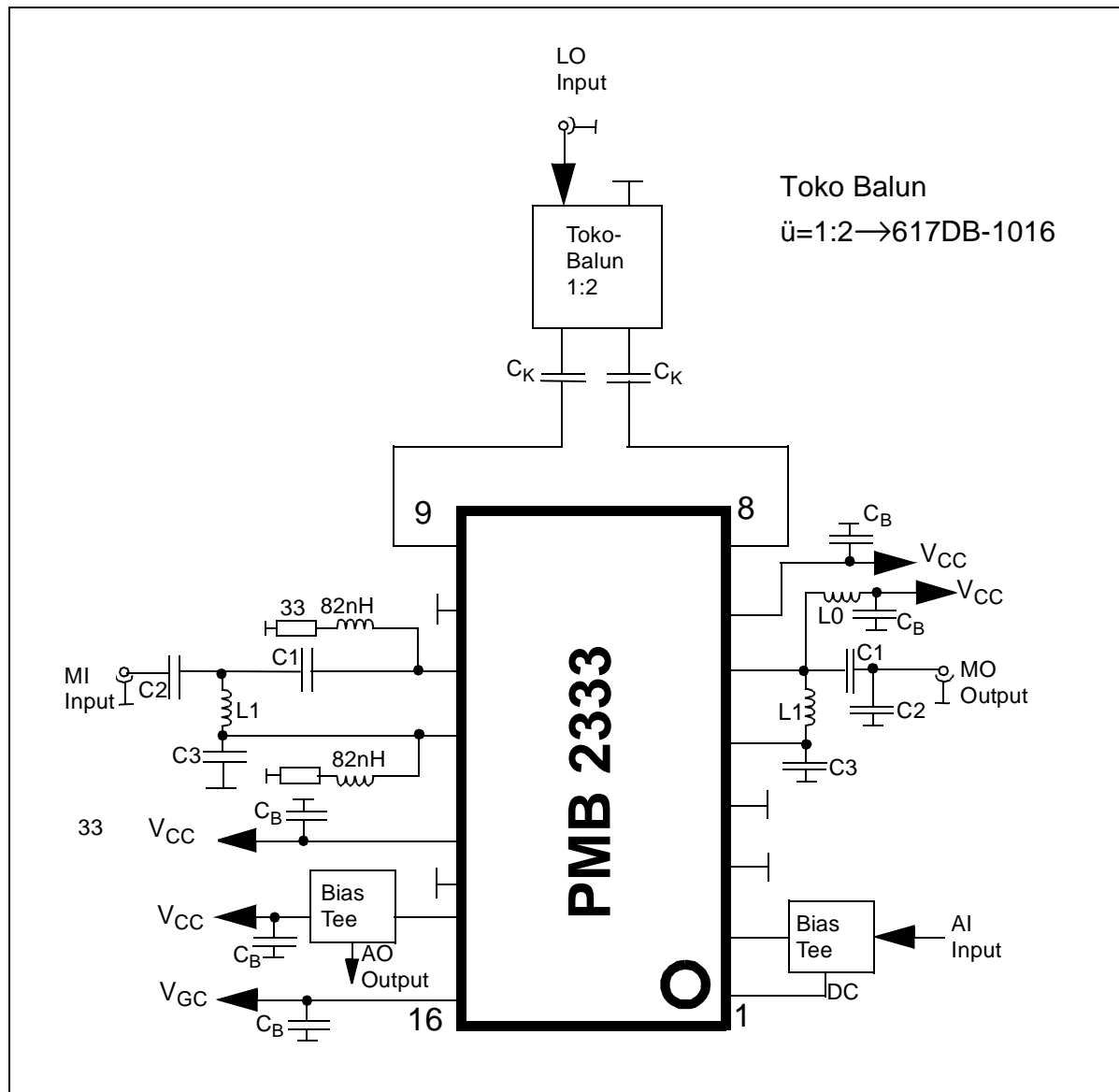
| | | | | | | | | |
|----|-----------------------|--------------|--|------|--|------|-------------|----|
| 58 | Output current | I_{MO+MOX} | | 4.0 | | mA | incl. R1,R2 | |
| 59 | Output resistance | R_{MODiff} | | 32 | | kOhm | IF=45MHz | 2a |
| 60 | | R_{MODiff} | | 25 | | kOhm | IF=300MHz | 2b |
| 61 | Output capacitance | C_{MODiff} | | 0.36 | | pF | IF=45MHz | 2a |
| 62 | | C_{MODiff} | | 0.39 | | pF | IF=300MHz | 2b |
| 63 | Power gain, IF=45MHz | P_{MI} | | 15 | | db | f=0.9GHz | 2a |
| 64 | | P_{MI} | | 14 | | db | f=1.8GHz | 2a |
| 65 | | P_{MI} | | 9 | | db | f=2.5GHz | 2a |
| 66 | Power gain, IF=300MHz | P_{MI} | | 7 | | db | f=0.9GHz | 2b |
| 67 | | P_{MI} | | 7 | | db | f=1.8GHz | 2b |
| 68 | | P_{MI} | | 2.5 | | db | f=2.5GHz | 2b |

MIXER, Isolation Between In-/Output, 0.9GHz, Downconversion, R1,2=180Ohm

| | | | | | | | | |
|----|----------|-------------|--|----|--|----|-------------------------------------|----|
| 69 | MI to MO | A_{MI-MO} | | 50 | | db | $f_{MI}=945MHz,$ $f_{LO}=900MHz$ | 2a |
| 70 | LO to MO | A_{LO-MO} | | 40 | | db | “ | 2a |
| 71 | LO to MI | A_{LO-MI} | | 35 | | db | “ | 2a |
| 72 | MO to MI | A_{MO-MI} | | 60 | | db | “ | 2a |
| 73 | MO to LO | A_{MO-LO} | | 60 | | db | “ | 2a |

2.4 Test Circuits

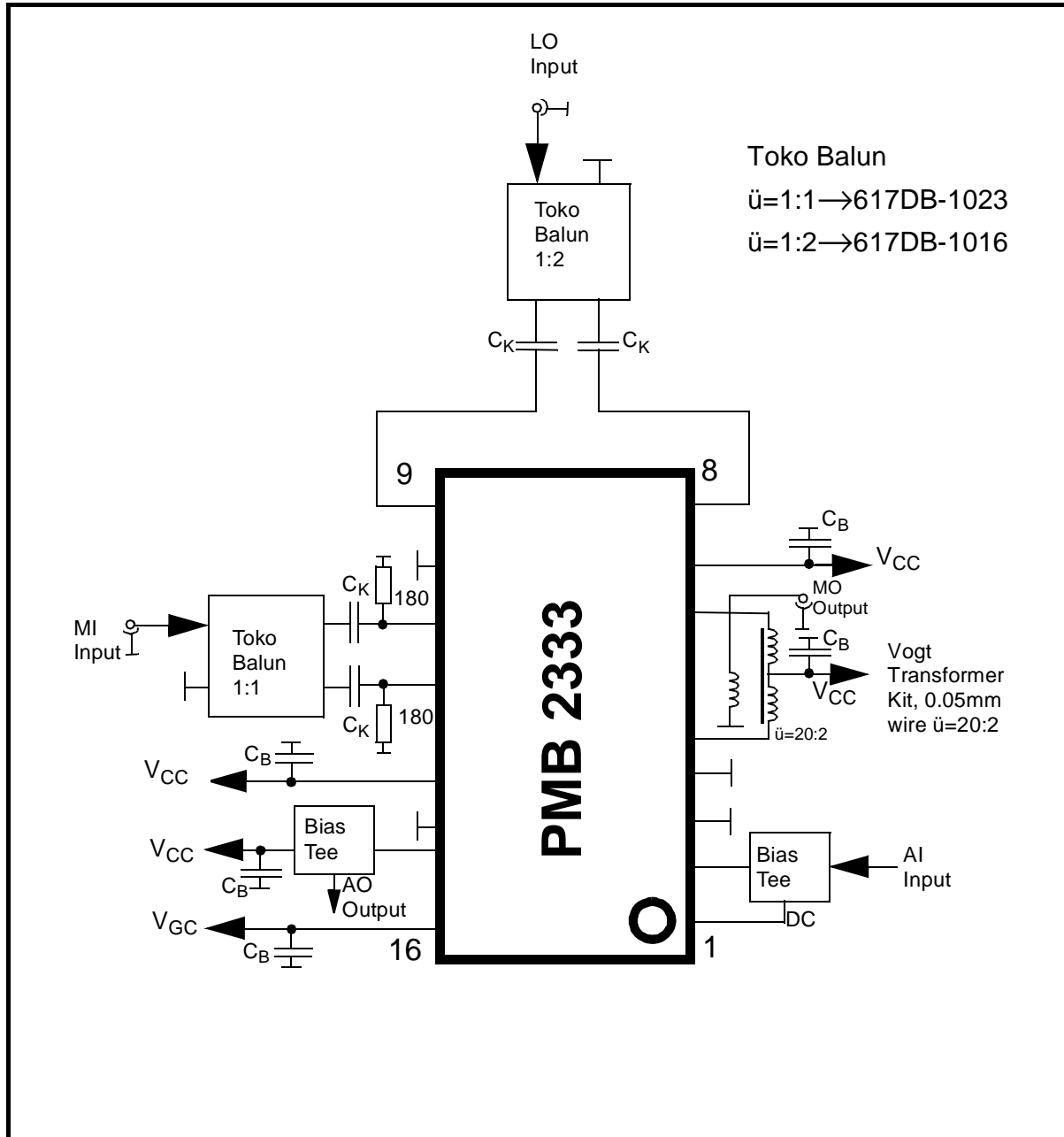
Test Circuit 1



Mixer/Driver Amplifier, Upconversion mode

| Test Circuit | f_{IF} [MHz] | L1[nH] | C1[pF] | C2[pF] | C3[pF] | C_K [pF] |
|--------------|----------------|--------|--------|--------|--------|------------|
| 1/MI | ≈660 | 8.2 | 4.7 | 56 | 10 | 15 |
| 1/MO | ≈2660 | 2.7 | 1.8 | 1.2 | 1 | X |

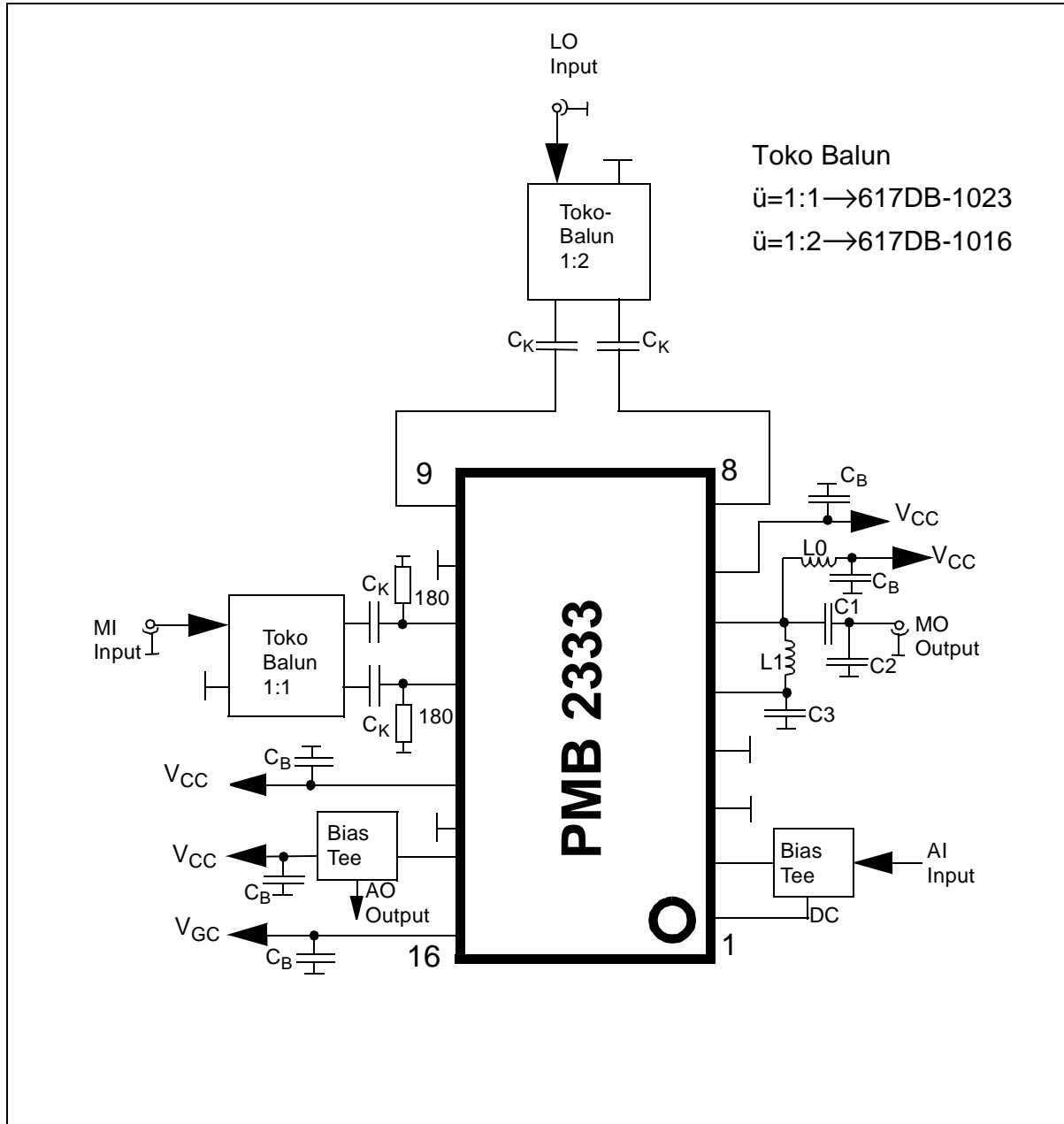
Test Circuit 2a



Mixer/Driver Amplifier, Downconversion mode

| Test Circuit | f_{IF} [MHz] | C_B [pF] | C_K [pF] | X | X |
|--------------|----------------|------------|------------|---|---|
| 2a | 45 | 15/100 | 15 | X | X |

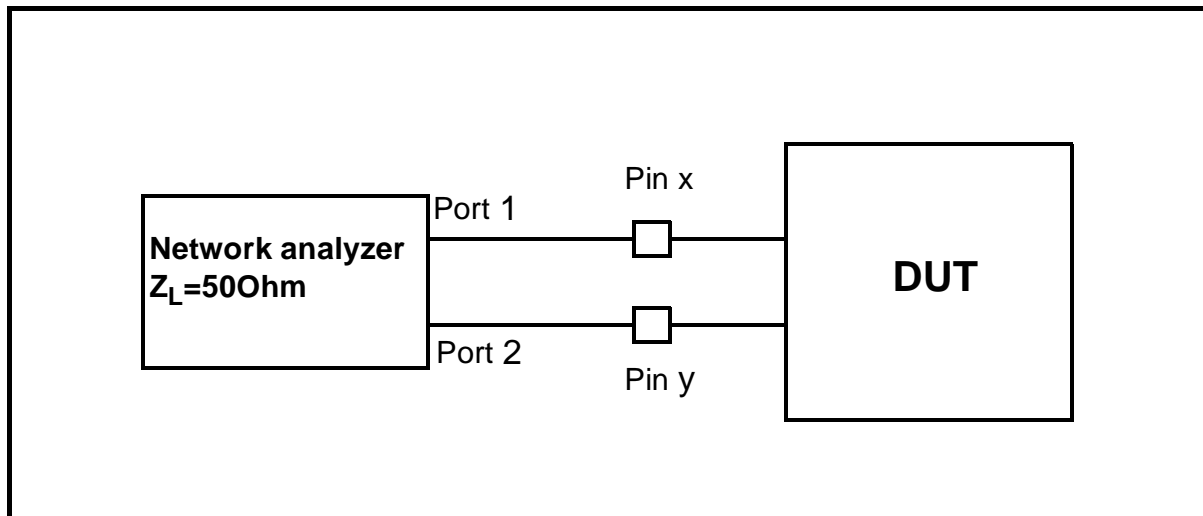
Test Circuit 2b



Mixer/Driver Amplifier, Downconversion mode

| Test Circuit | f_{IF} [MHz] | L0[nH] | L1[nH] | C1[pF] | C2[pF] | C3[pF] | Ck[pF] |
|--------------|----------------|--------|--------|--------|--------|--------|--------|
| 2b | ≈300 | 680 | 150 | 2.7 | 12 | 1.8 | 15p |

Test Circuit 3



**S-Parameter Measurement of Amplifier
S11, S12, S21, S22**

The S-Parameters are tested at the indicated frequency on Duroid 5880 Teflon Boards.

Via the NWA the capacitive coupling is done.
The output levels at port1 and 2 for pin x and y are -30dbm.

S11 and S22 have to be considered as design hints and are measured with SIEMENS testboards.

| Test | Test frequency MHz | Pin X | Pin Y |
|---------------------------|-------------------------------|--------------|--------------|
| Amp.S11, S12, S21, S22 | 30 - 3000 | AI | AO |

Diagram 3a
 S-Parameter Amplifier $I_{AO}=20\text{mA}$, $V_{CC}=3.3\text{V}$, $f=30\text{-}3000\text{MHz}$

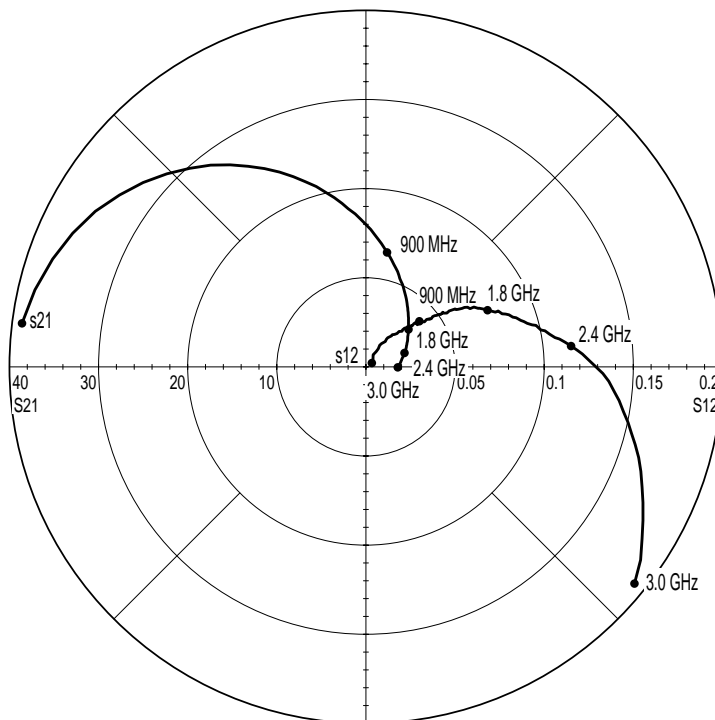
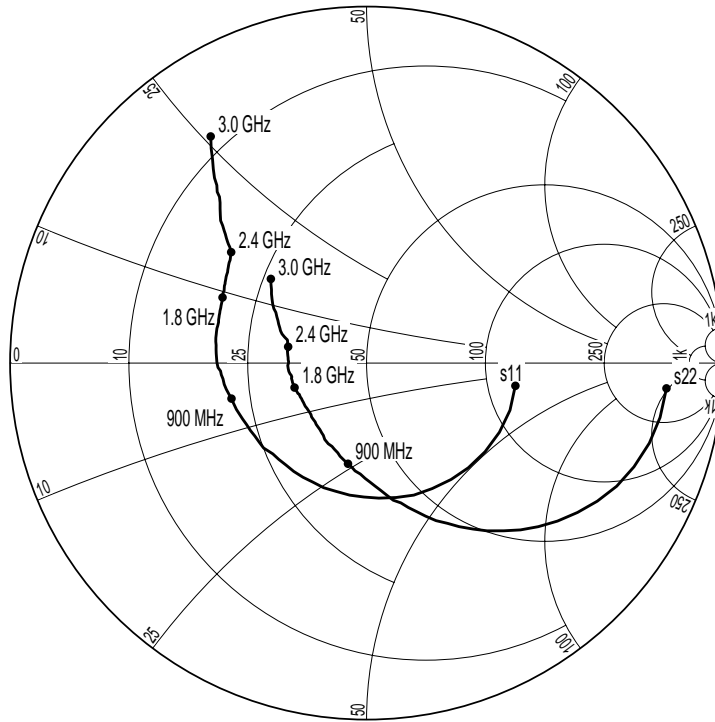
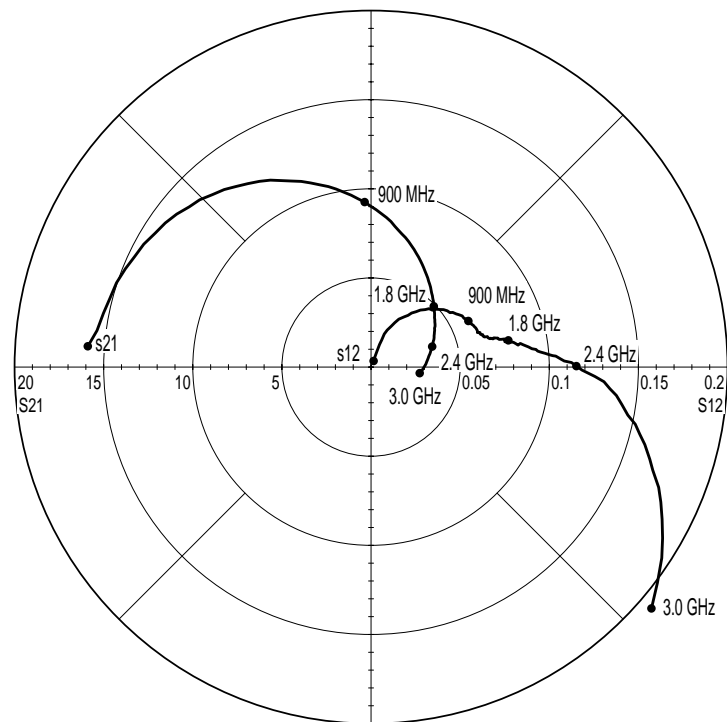
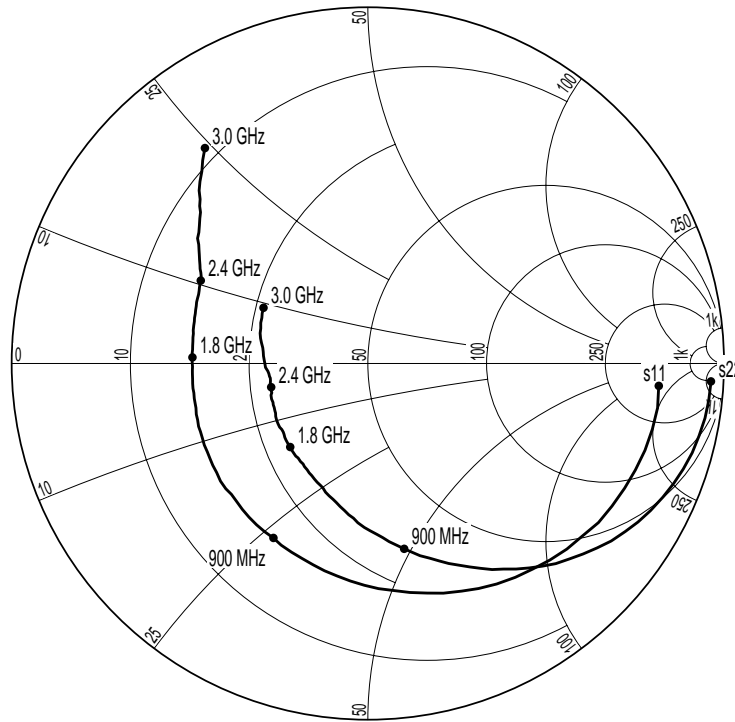


Diagram3b

S-Parameter Amplifier $I_{AO}=5mA$, $V_{CC}=3.3V$, $f=30-3000MHz$



S-Parameter Amplifier $I_{AO}=0mA$, $V_{CC}=3.3V$, $f=30-3000MHz$

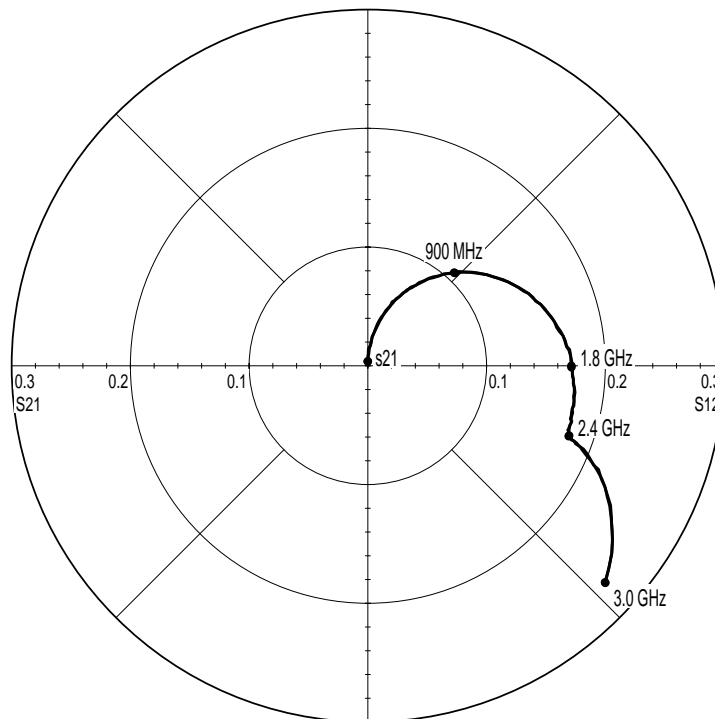
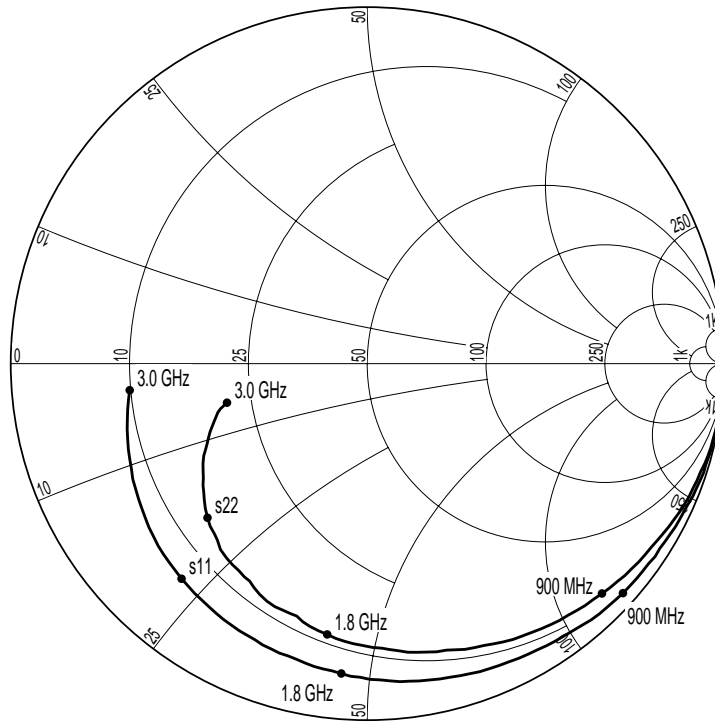
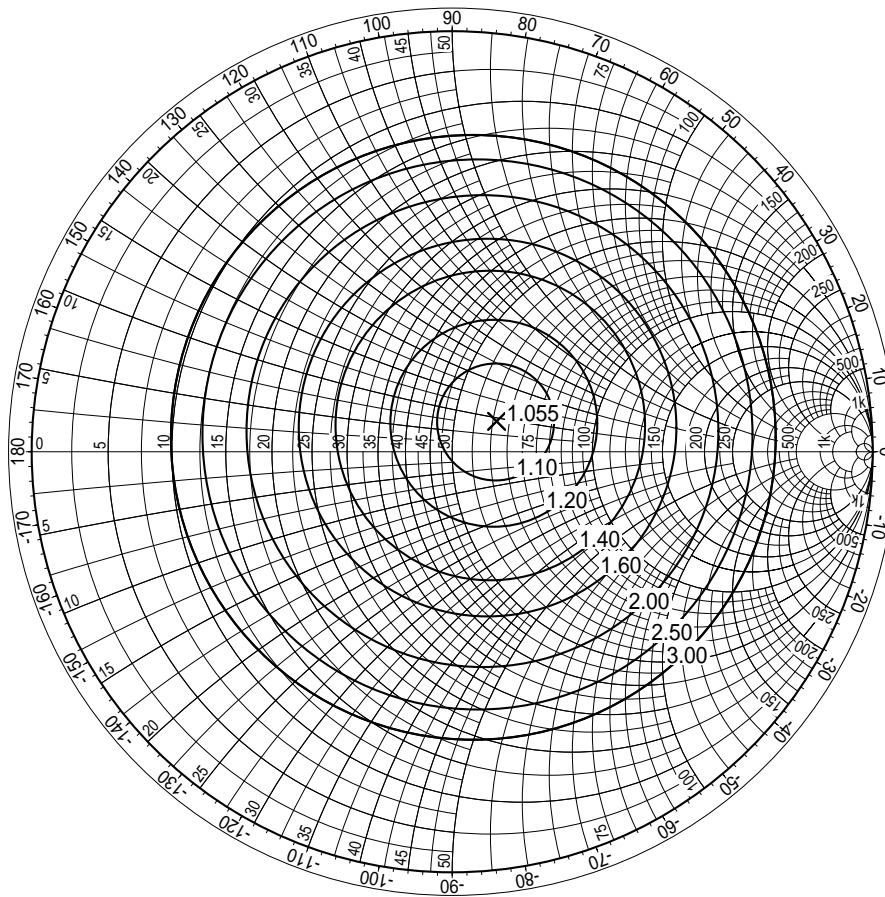


Diagram 3d
Noise Circles Amplifier $I_{AO}=5mA$, $V_{CC}=3.3V$; $f=0.9GHz$



$F_{MIN} = 1.055dB$

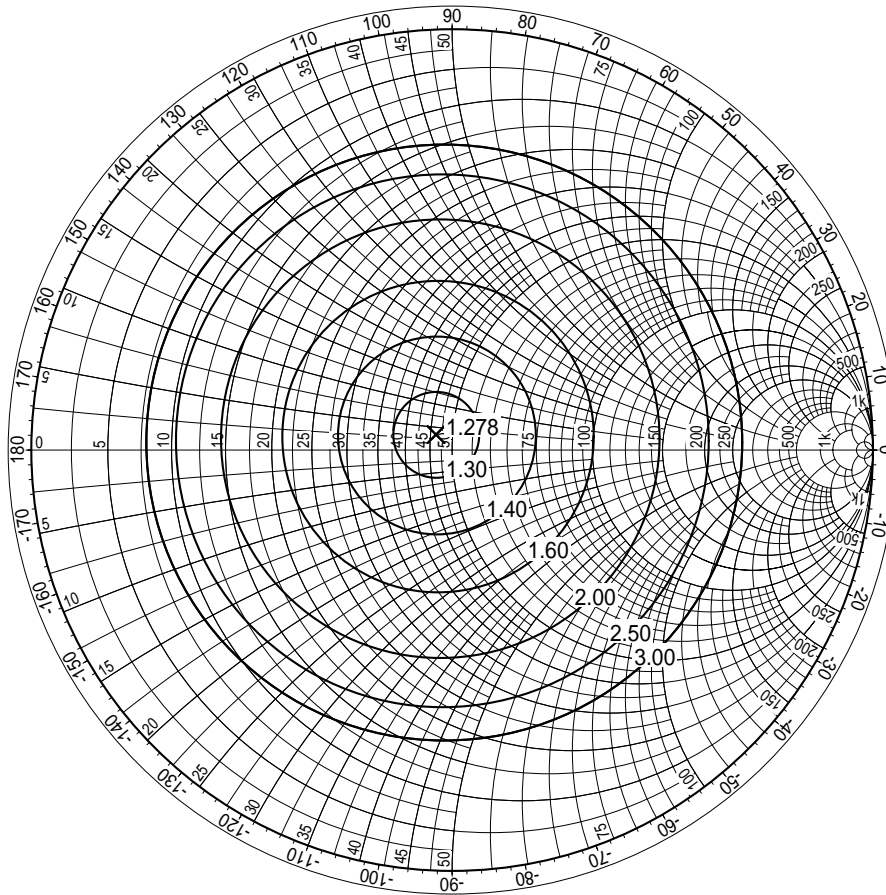
$R_n = 10.17\Omega$

$G_{OPT} = 16.01mS$

$B_{OPT} = -2.36mS$

$\Gamma_{OPT} = 0.128 \exp(j 34.4^\circ)$

Diagram 3e
Noise Circles Amplifier $I_{AO}=5mA, V_{CC}=3.3V; f=1.8GHz$



$F_{MIN} = 1.278dB$

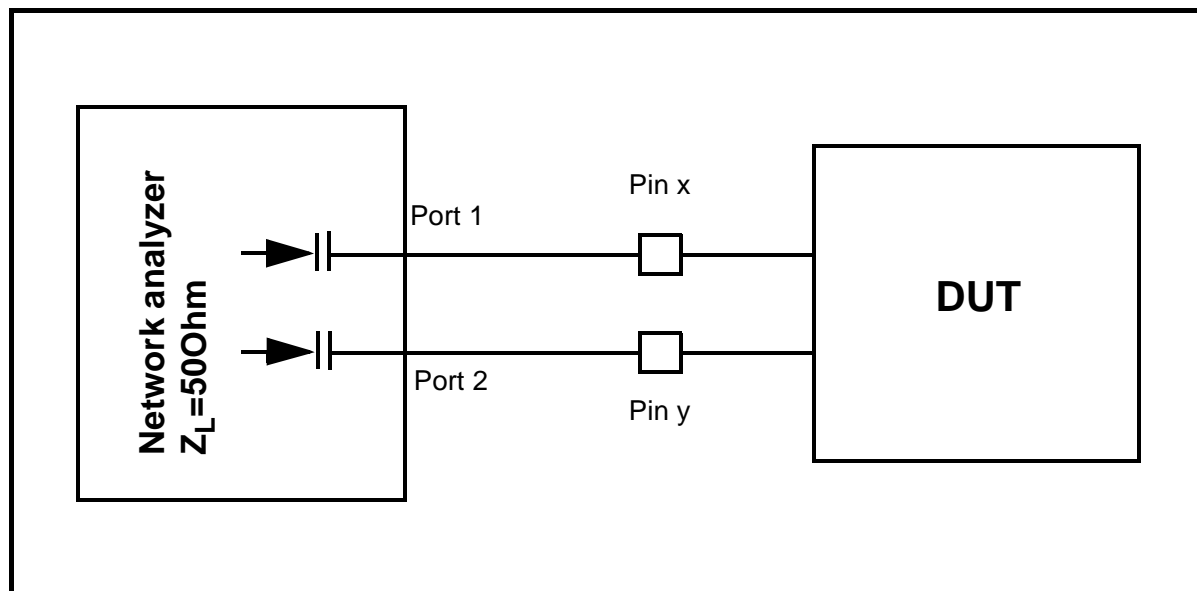
$R_n = 7.52\Omega$

$G_{OPT} = 21.54mS$

$B_{OPT} = -1.59mS$

$\Gamma_{OPT} = 0.053 \exp(j 136.2^\circ)$

Test Circuit 4



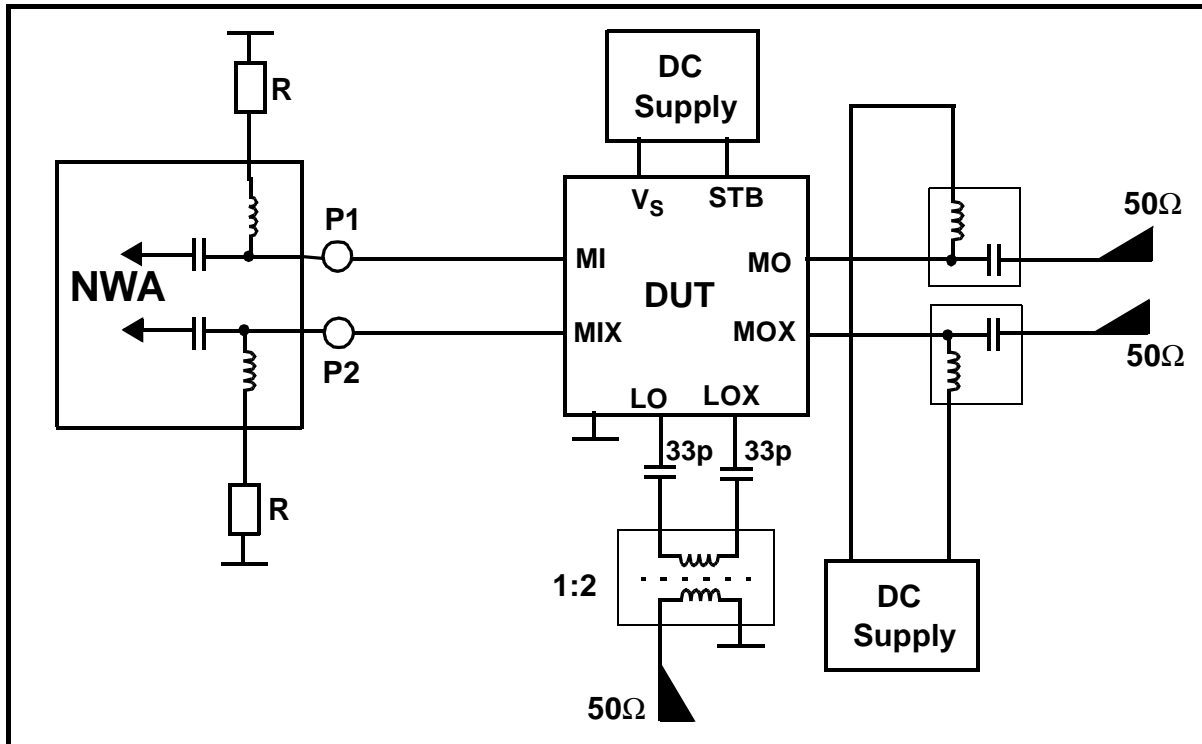
S-Parameter Measurement of Mixer
S11, S12, S21, S22

| Test | Test Frequency [MHz] | Pin X | Pin Y |
|---------------------|----------------------|-------|-------|
| LO-Input impedance | 30 - 3000 | 8 | 9 |
| Mi-Input impedance | 30 - 3000 | 11 | 12 |
| MO-Output impedance | 30 - 3000 | 5 | 6 |

The S-Parameters are tested at the indicated frequency and the equivalent parallel or series circuit is calculated on this base.

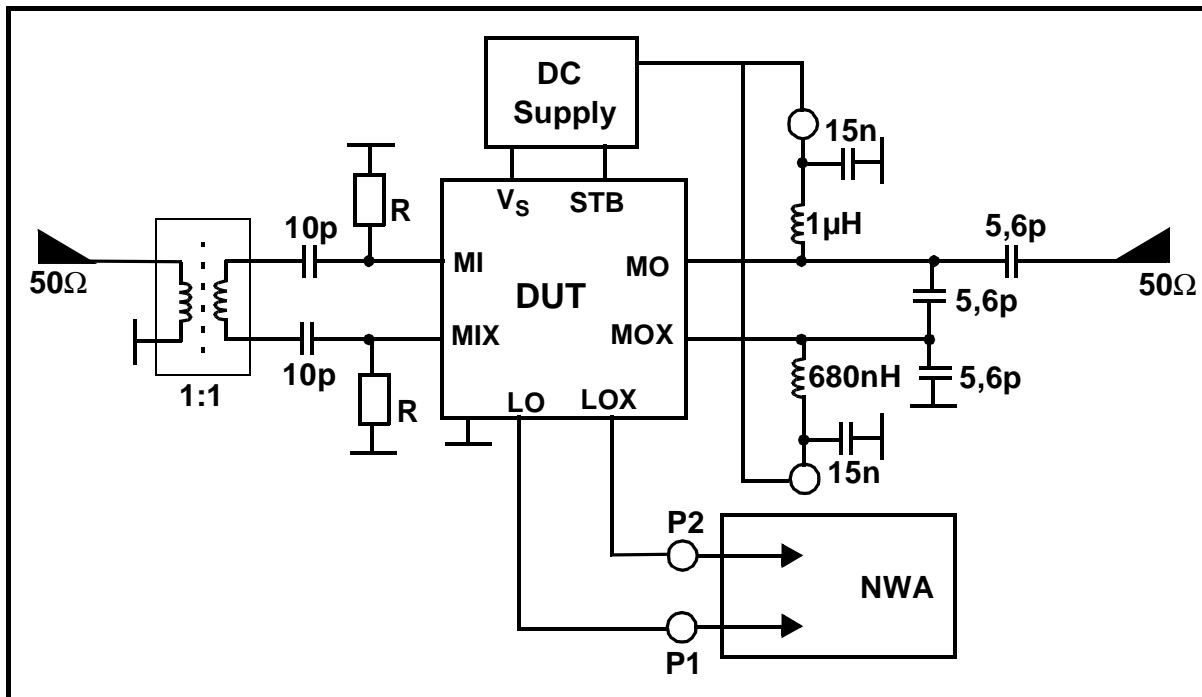
Via the NWA the capacitive coupling is done and the open collector pins are connected to VCC. The output levels at port1 and 2 for pin x and y are -30dbm for MI and MO-impedances and -5dbm for the LO impedance. S-Parameters have to be considered as design hints and are measured with SIEMENS testboards.

Test Circuit 4a



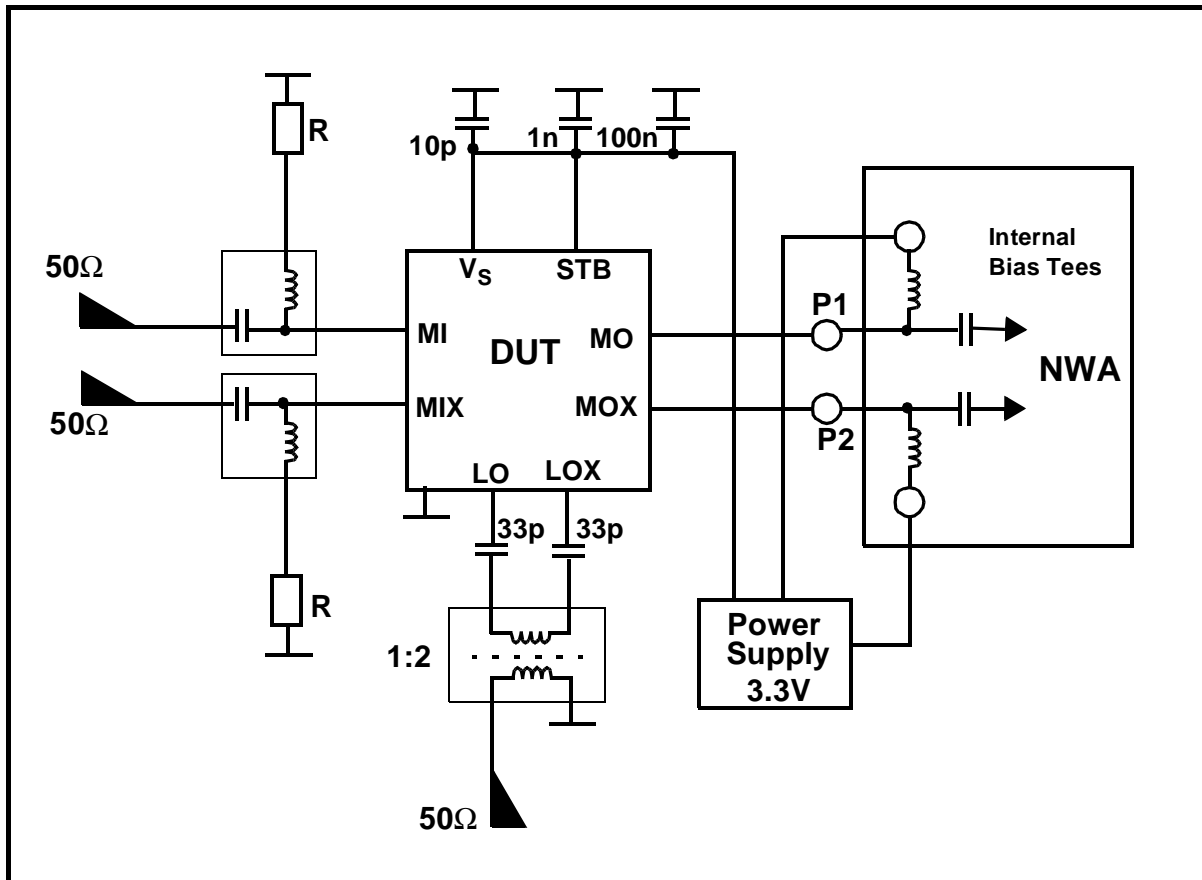
Mixer Input Impedance Measurement

Test Circuit 4b



Mixer Local Oscillator Impedance Measurement

Test Circuit 4c



Mixer Output Impedance Measurement

Diagram 4a
Mixer MI Input Impedance Z_{MI} , $I_{MO/MOX} = 10\text{mA}$, $f=30\text{-}3000\text{MHz}$

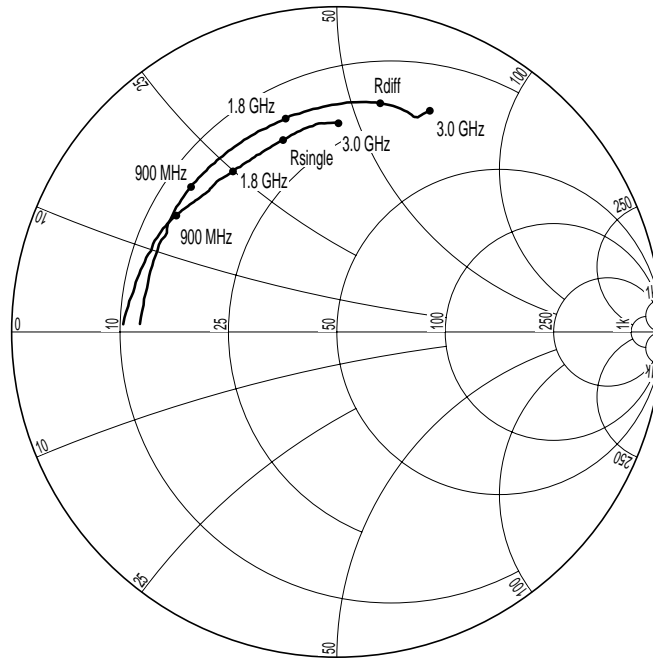


Diagram 4b
Mixer MI Input Impedance Z_{MI} , $I_{MO/MOX} = 4\text{mA}$, $f=30\text{-}3000\text{MHz}$

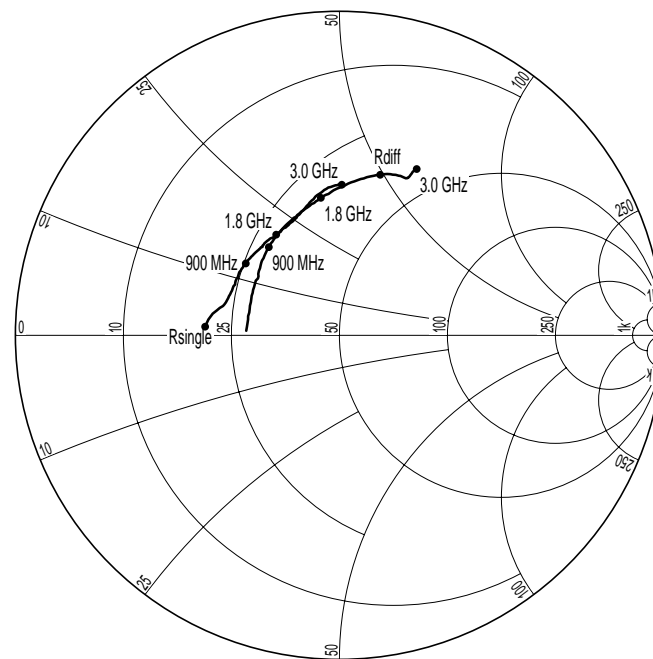


Diagram 4c
Mixer LO Input Impedance Z_{LO} , $I_{MO/MOX} = 10mA$, $f=30-3000MHz$

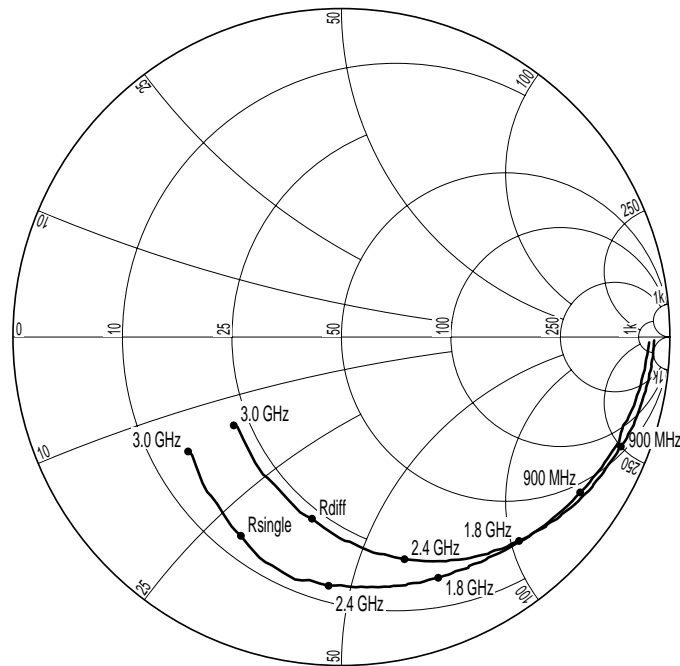


Diagram 4d
Mixer LO Input Impedance Z_{LO} , $I_{MO/MOX} = 4mA$, $f=30-3000MHz$

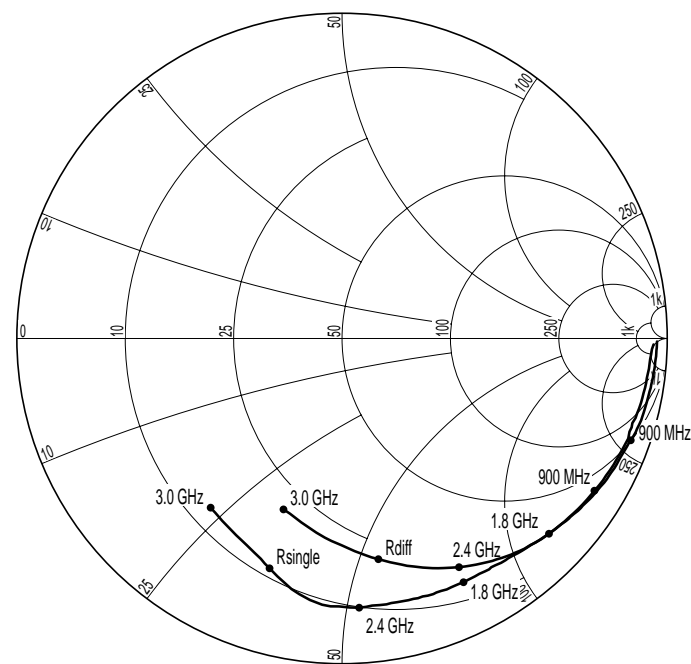
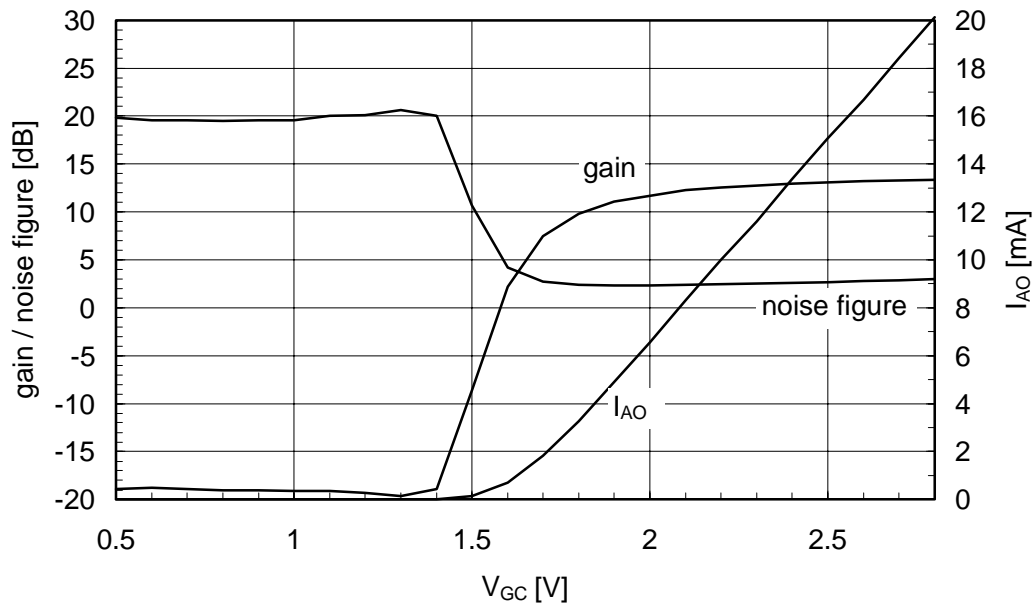


Diagram 5



Gain, Noise Figure and I_{AO} versus Gain Control voltage
 Noise Figure values without correction of attenuation (0.4 dB) at input of the amplifier -> $NF_{min}=1.7dB$ at $V_{GC}=1.91V$, amplifier current $I_{AO}=5mA$, open collector voltage $V_{AO}=3.3V$ (according test circuit 2, $f=1.8GHz$).

3 Application Data

3.1 Receiver Application

3.1.1 Shortform Data

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 2.7\text{ V}$

LNA and Mixer input signal $f_{RF} = 1960\text{ MHz}$, $P_{RF} = -30\text{ dBm}$

LO signal $f_{LO} = 1735\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

IF output $f_{IF} = 225\text{ MHz}$

All measurements refer to SMA connectors without consideration of PCB losses

| Parameter | Symbol | Limit values | | | Unit | Remarks |
|---------------------------------|---------------|--------------|------|------|------|--|
| | | min. | typ. | max. | | |
| Mixer section | | | | | | |
| Mixer current | I_{Mixer} | | 4 | | mA | $I_{MO} + I_{MOX}$ |
| Conversion gain | G_C | | 8.5 | | dB | |
| Noise Figure (DSB) | NF_{DSB} | | 6.4 | | dB | |
| 3rd order input intercept point | $IICP3$ | | +1 | | dBm | |
| 1dB-compression point | P_{1dB} | | -8 | | dBm | |
| Input blocking level | P_{BL} | | -7 | | dBm | $\Delta f = 800\text{ kHz}$, -3dB for wanted signal |
| Port matching | | | | | | |
| RF return loss | $ S_{11,RF} $ | | 12 | | dB | |
| LO return loss | $ S_{11,LO} $ | | 10 | | dB | |
| IF return loss | $ S_{11,IF} $ | | 11 | | dB | |
| Isolations | | | | | | |
| LO to IF output | A_{LO-IF} | | 40 | | dB | $f = 1735\text{ MHz}$ |

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 2.7\text{ V}$

LNA and Mixer input signal $f_{RF} = 1960\text{ MHz}$, $P_{RF} = -30\text{ dBm}$

LO signal $f_{LO} = 1735\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

IF output $f_{IF} = 225\text{ MHz}$

All measurements refer to SMA connectors without consideration of PCB losses

| Parameter | Symbol | Limit values | | | Unit | Remarks |
|---------------------------------|---------------|--------------|------|------|------|---|
| | | min. | typ. | max. | | |
| LO to RF input | A_{LO-RF} | | 43 | | dB | $f_{LO} = 1735\text{ MHz}$ $f_{image,min} = 1480\text{ MHz}$ $f_{signal,max} = 1990\text{ MHz}$ |
| | | | 25 | | dB | |
| | | | 35 | | dB | |
| RF input to LO | A_{RF-LO} | | 48 | | dB | $f = 1960\text{ MHz}$ |
| LNA section | | | | | | |
| LNA current | I_{LNA} | | 4.6 | | mA | |
| Gain | G | | 12.5 | | dB | |
| Noise Figure | NF | | 1.8 | | dB | assuming a PCB loss of 0.25 dB at AI, typical noise figure at matching circuit results to 1.55 dB |
| 3rd order input intercept point | IICP3 | | 0 | | dBm | |
| 1dB-compression point | P_{1dB} | | -10 | | dBm | |
| AI return loss | $ S_{11,AI} $ | | 11 | | dB | |
| AO return loss | $ S_{11,AO} $ | | 14 | | dB | |

3.1.2 Measurement results

Mixer section

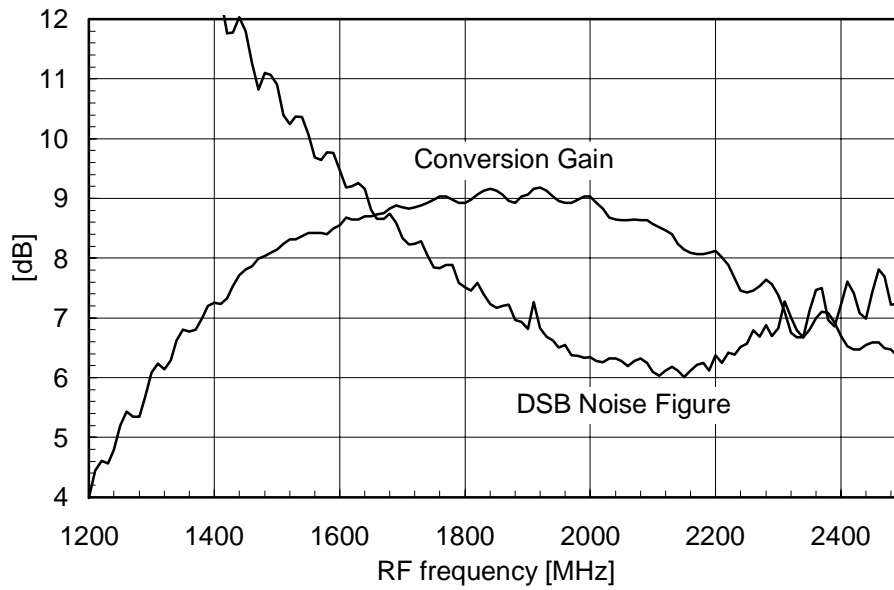


Figure 1: Conversion Gain an Noise Figure versus Frequency

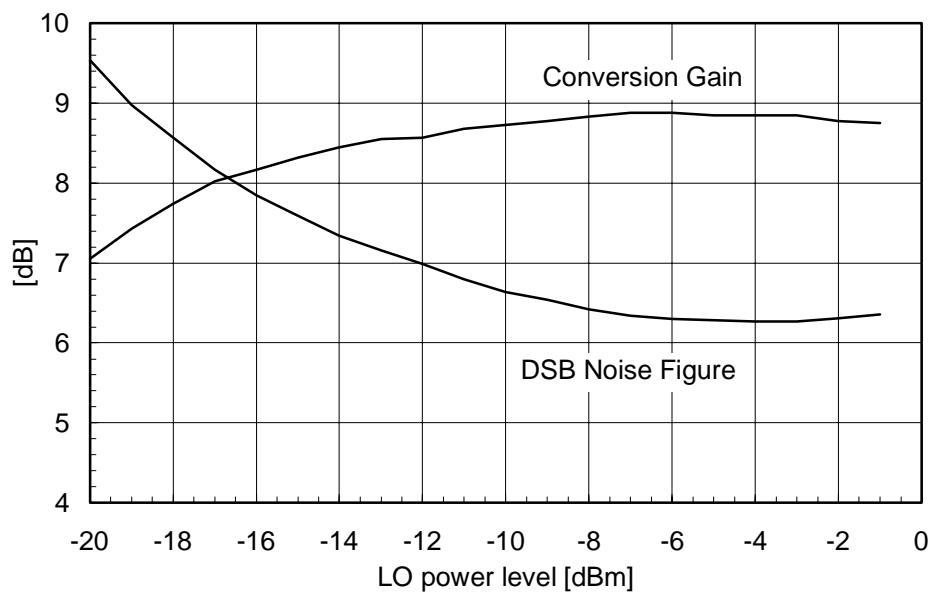


Figure 2: Conversion Gain and Noise Figure versus LO power

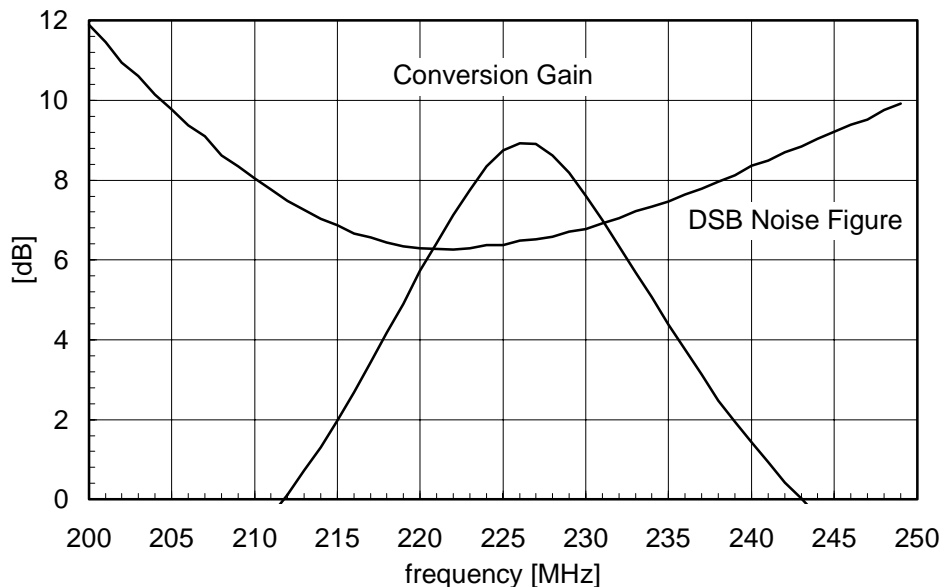


Figure 3: Conversion Gain and Noise Figure versus IF frequency

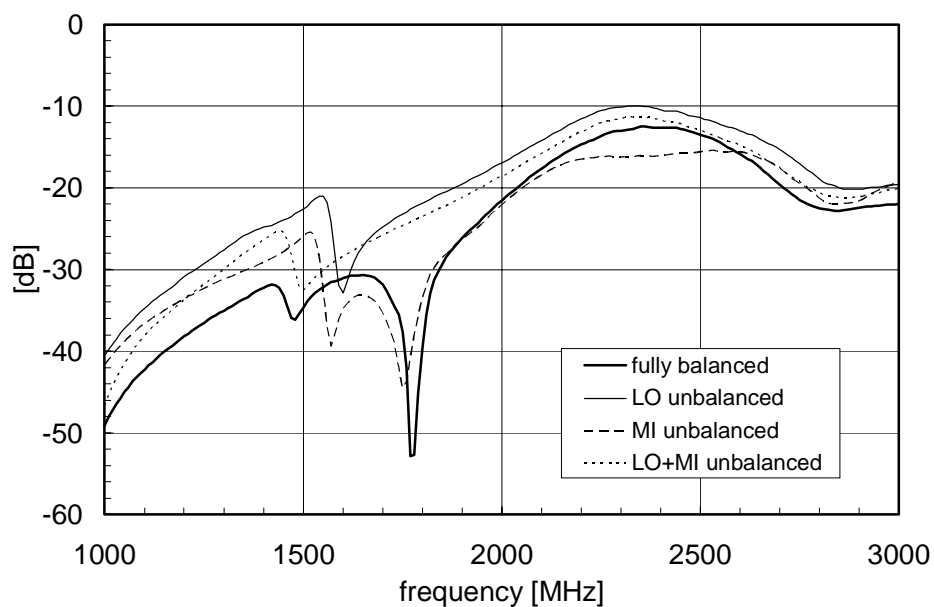


Figure 4: Isolation LO to IF

In unbalanced case the matching network is replaced by a 10pF series capacitor to one port pin. The other port pin is tied to ground via a second 10pF capacitor. This means **no** power matching is done.

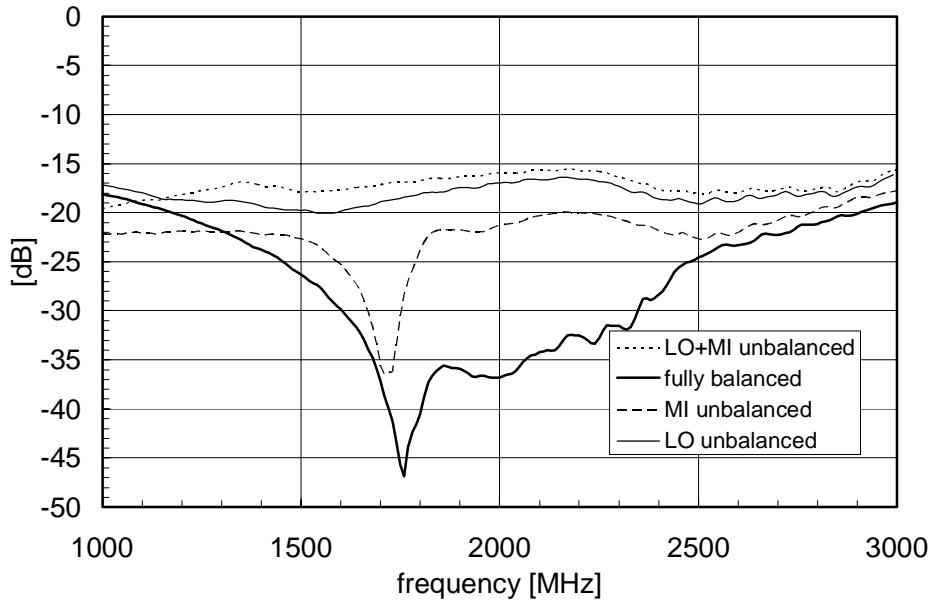


Figure 5: Isolation LO to RF

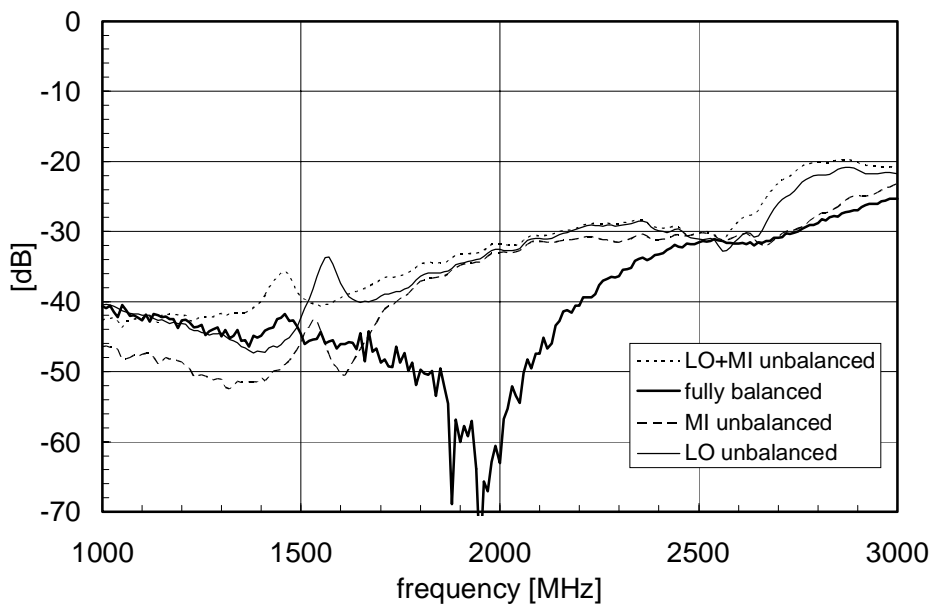


Figure 6: Isolation RF to LO

LNA section

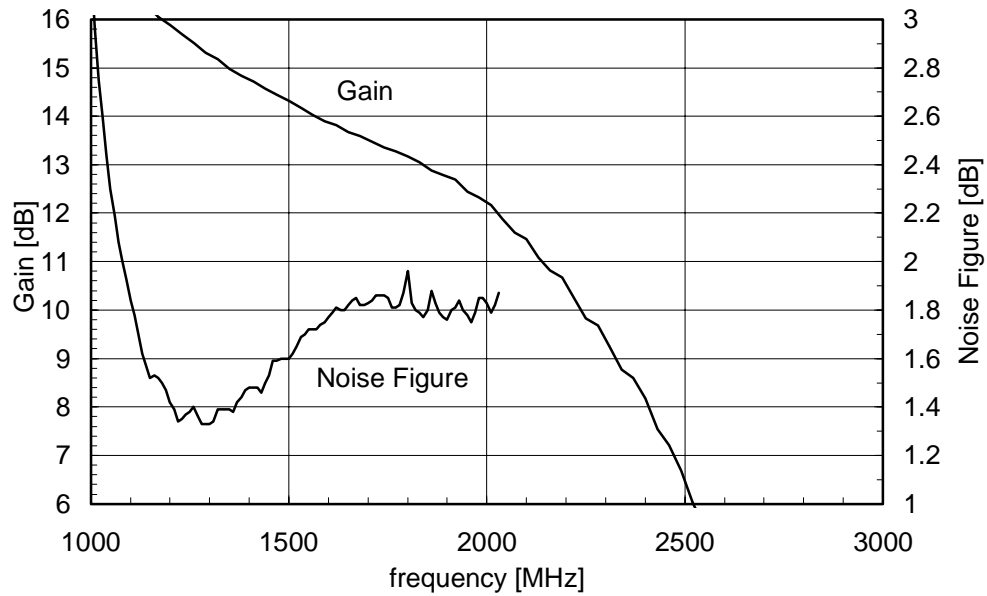


Figure 7: Gain and noise figure versus frequency

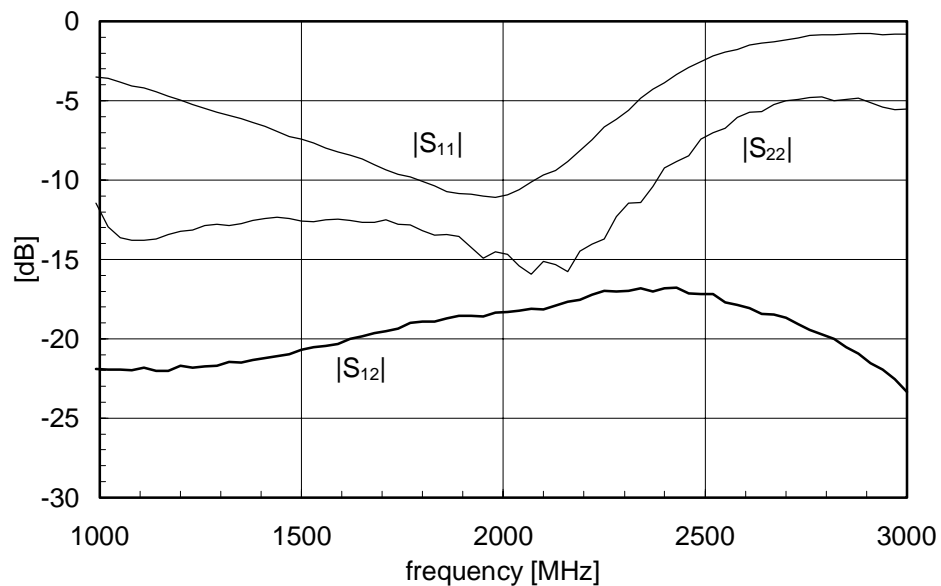


Figure 8: Reverse isolation and return loss versus frequency

3.1.3 Application hint - Mixer metrics versus mixer current

Mixer current may be increased to obtain a higher Input 3rd Order Intercept (IIP3), higher 1 dB Compression Point (P1dB), and increased Conversion Gain (G). In a typical application, in order to increase mixer current from the minimum level of 800µA, R5 and R6 are added as shown in the circuit diagram at the end of this section. These external resistors are placed in parallel to existing bias resistors internal to the PMB2333, thereby reducing the aggregate resistance in the emitters and increasing current. As current is increased by further reduction of the value of R5 and R6, the mixer inputs MI and MIX may begin to suffer ‘RF Loading’ unless RF chokes are used between the MI/MIX pins and R5 / R6 (compare test circuit 1).

For the data presented in this section, mixer current was varied in a different manner. To eliminate the effects of RF impedance variation (due to ‘RF Loading’) at MI / MIX caused by changing the values of R5 and R6 for different mixer currents, these resistors were set equal to 1kΩ. Mixer current was then varied by adjusting the power supply voltage Vx. Note that Vx may take on positive values with respect to ground for low (e.g. 1mA) mixer currents, or negative values for higher currents.

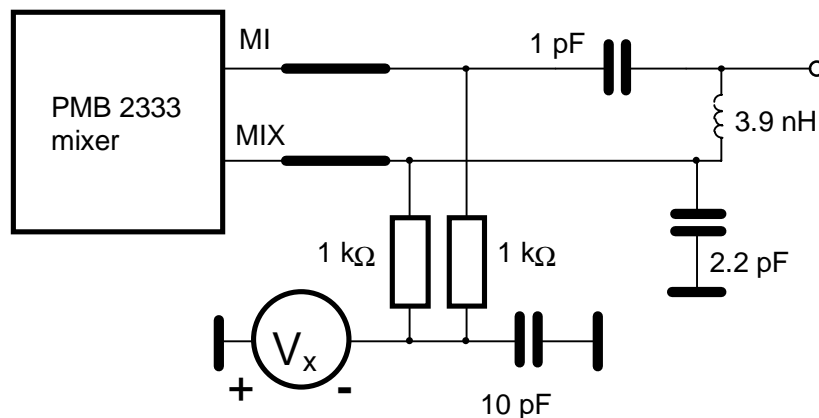


Figure 9: Modified mixer input circuitry

Note that the mixer input impedance seen at MI / MIX is a strong function of mixer current. The mixer input balanced-to-unbalanced transformer/matching circuitry was originally tuned for a current of 4mA, and was **not** re-optimized for each of the other current levels. Despite this limitation, the return loss at the 50Ω port is better than 10dB over the entire 1 to 10 mA current range. The mixer output (MO / MOX) and local oscillator (LO / LOX) ports exhibit negligible change in impedance over this same current range.

Measurement conditions: $T_A = 25\text{ °C}$
 $V_{CC} = 2.7\text{ V and } 4.5\text{V}$
 $f_{RF} = 1960\text{ MHz}$
 $f_{LO} = 1735\text{ MHz, } P_{LO} = -6\text{ dBm}$

The Effect of Power Supply Voltage on Mixer Metrics

In seeking to improve Mixer Input Third-Order Intercept and 1dB Compression Point, it is important to understand the constraints on these parameters imposed by power supply voltage. **Refer to Figure 10.**

Receiver 'Blocking' is predominantly influenced by the Mixer's 1 dB Compression Point (P1dB) and **not** the Input 3rd Order Intercept Point (IIP3). For a supply voltage of 4.5V, mixer P1dB (referred to the input of the application circuitry) increases with additional mixer current, and begins to flatten out above 8mA. When supply voltage is decreased to 2.7V, mixer P1dB starts flattening out at around 5mA.

Note how, at the 2.7V supply voltage, IIP3 continues to increase at currents over 6mA while P1dB flattens out. For a given conversion gain G, while operating at 2.7 Volts, considering only the mixer's Input 3rd Order Intercept (IIP3) might lead one to falsely conclude that increasing current beyond 6mA improves receiver blocking. If, however, conversion gain is decreased, it may be possible to improve the receiver's blocking level with additional mixer current.

As shown in Figure 10, the 1 dB compression level, referred to the input, is limited by either current or the available voltage swing at the mixer output. The transition between these two regions takes place at 4mA for the specified conversion gain and supply voltage.

Figure 11 gives the DSB mixer noise figure versus the mixer current.

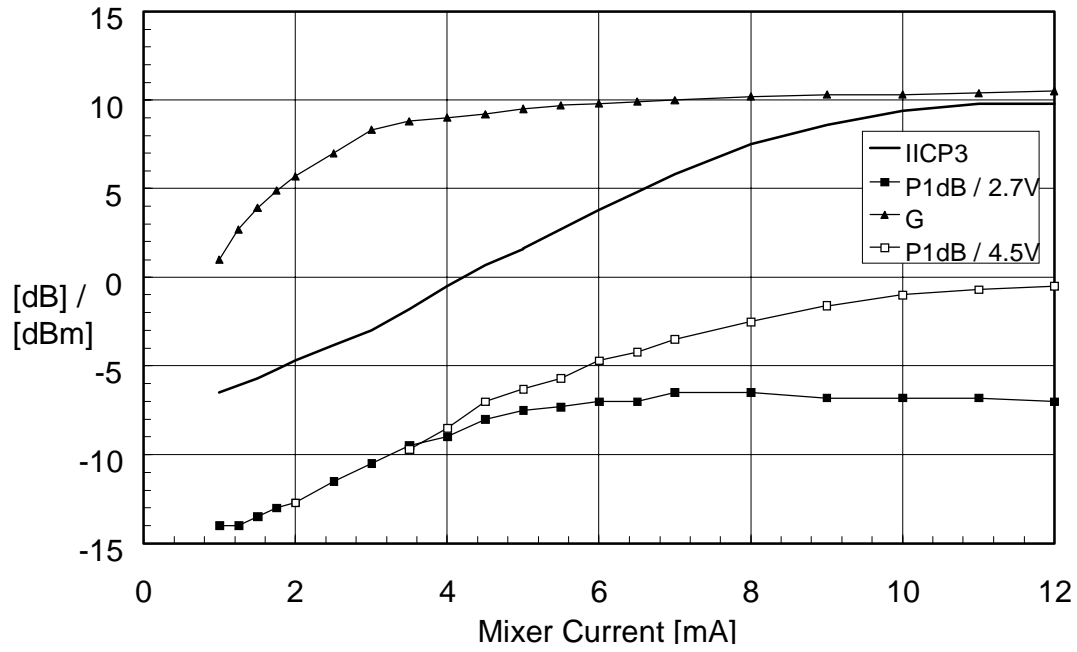


Figure 10: Mixer Input Intercept Point (IIP3), 1 dB Compression Point (P1dB) Referred to Input, Gain (G).

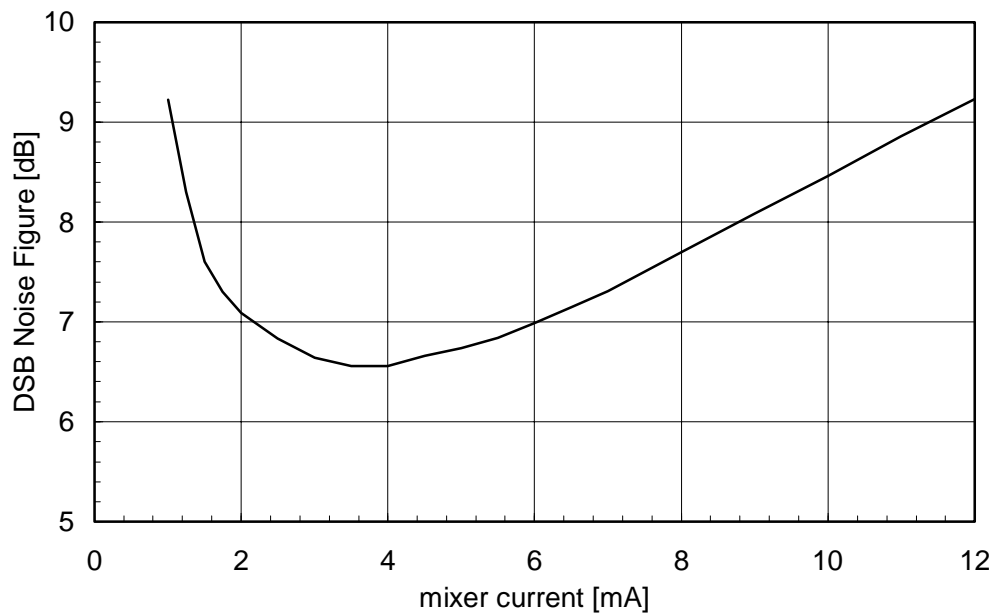


Figure 11: Mixer DSB Noise Figure

3.1.4 Circuit diagram and PCB layout

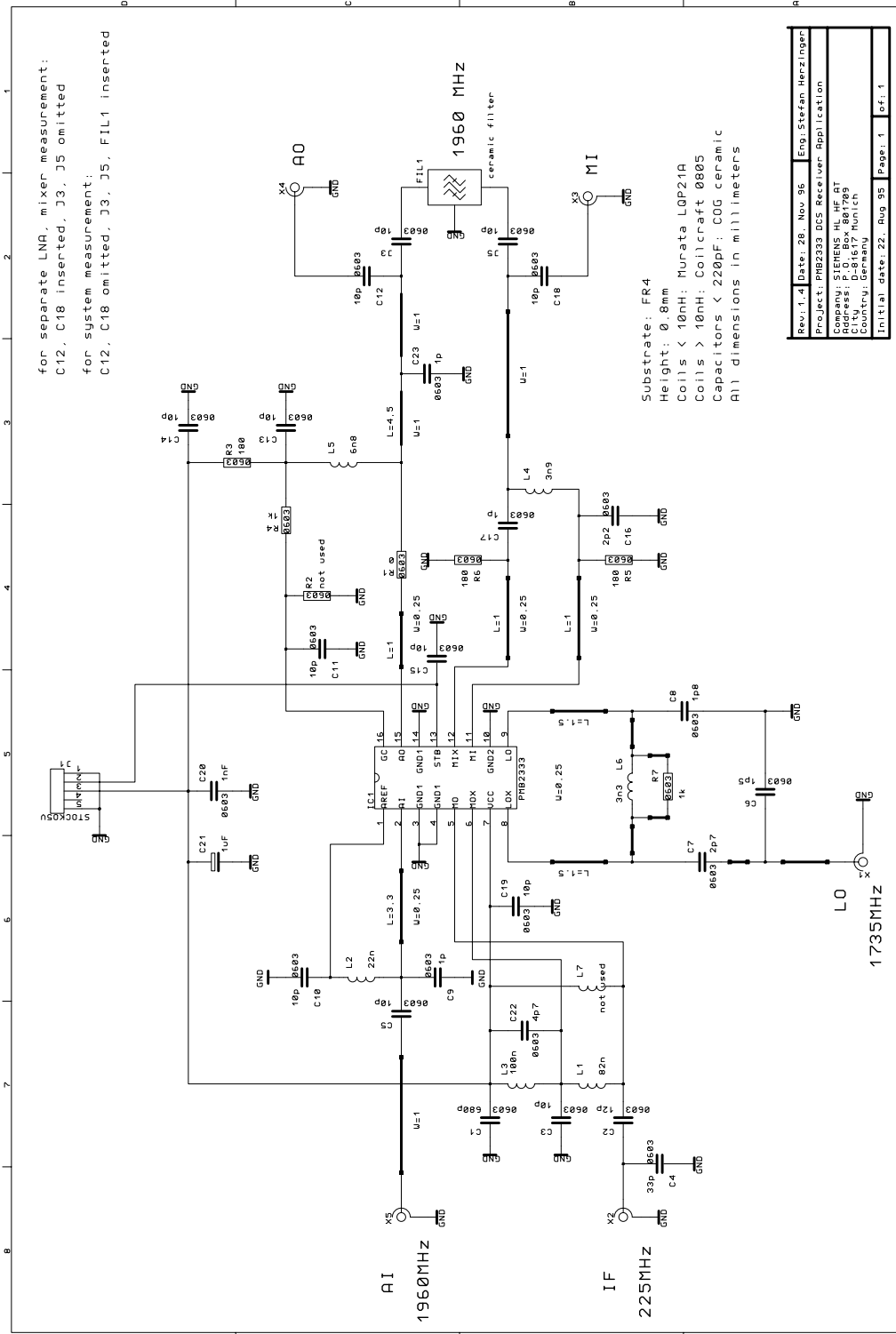


Figure 12: Circuit diagram

PCB dimensions: 80 x 50 mm

Substrate material: FR4

Substrate height: 0.8 mm

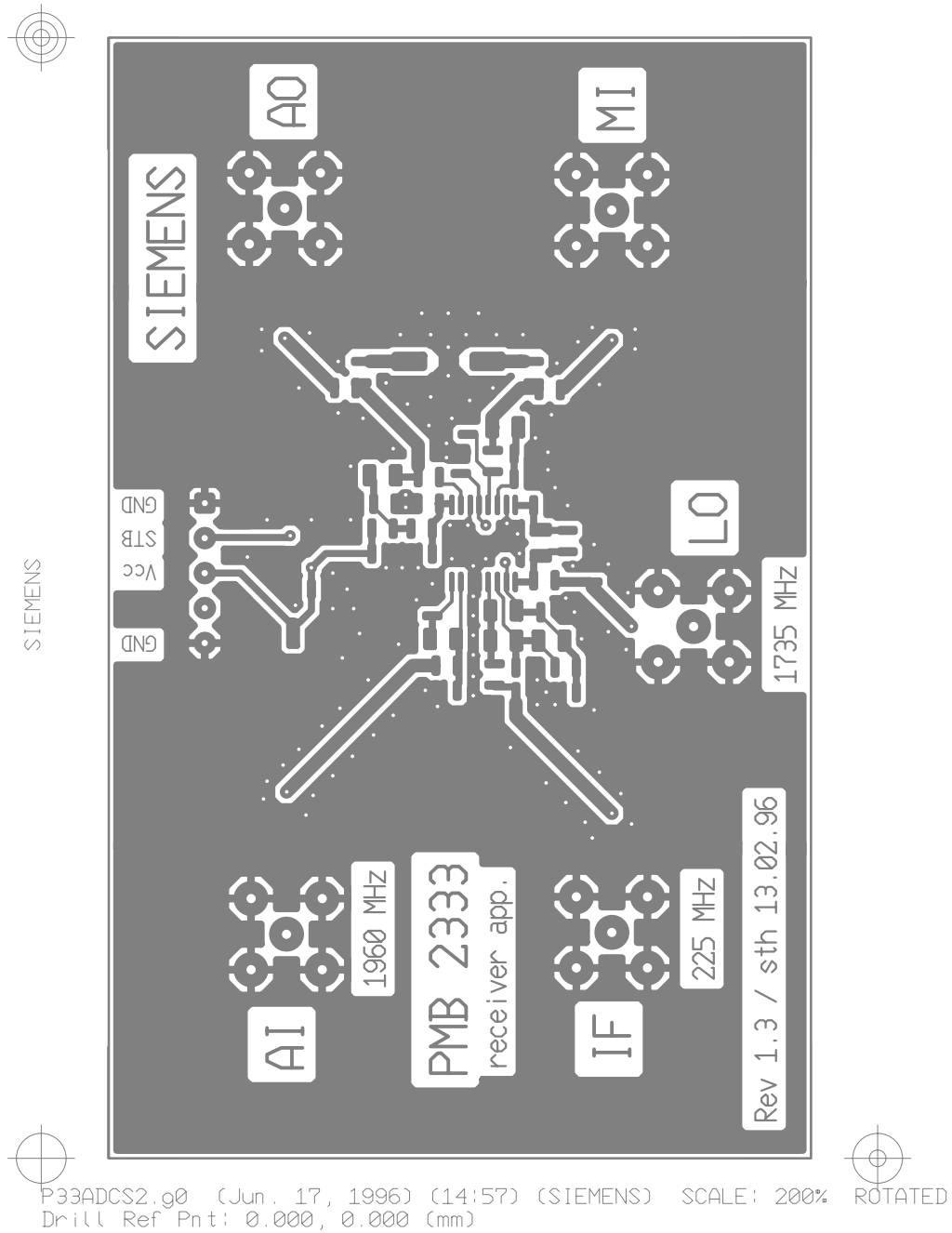


Figure 13: PCB top side

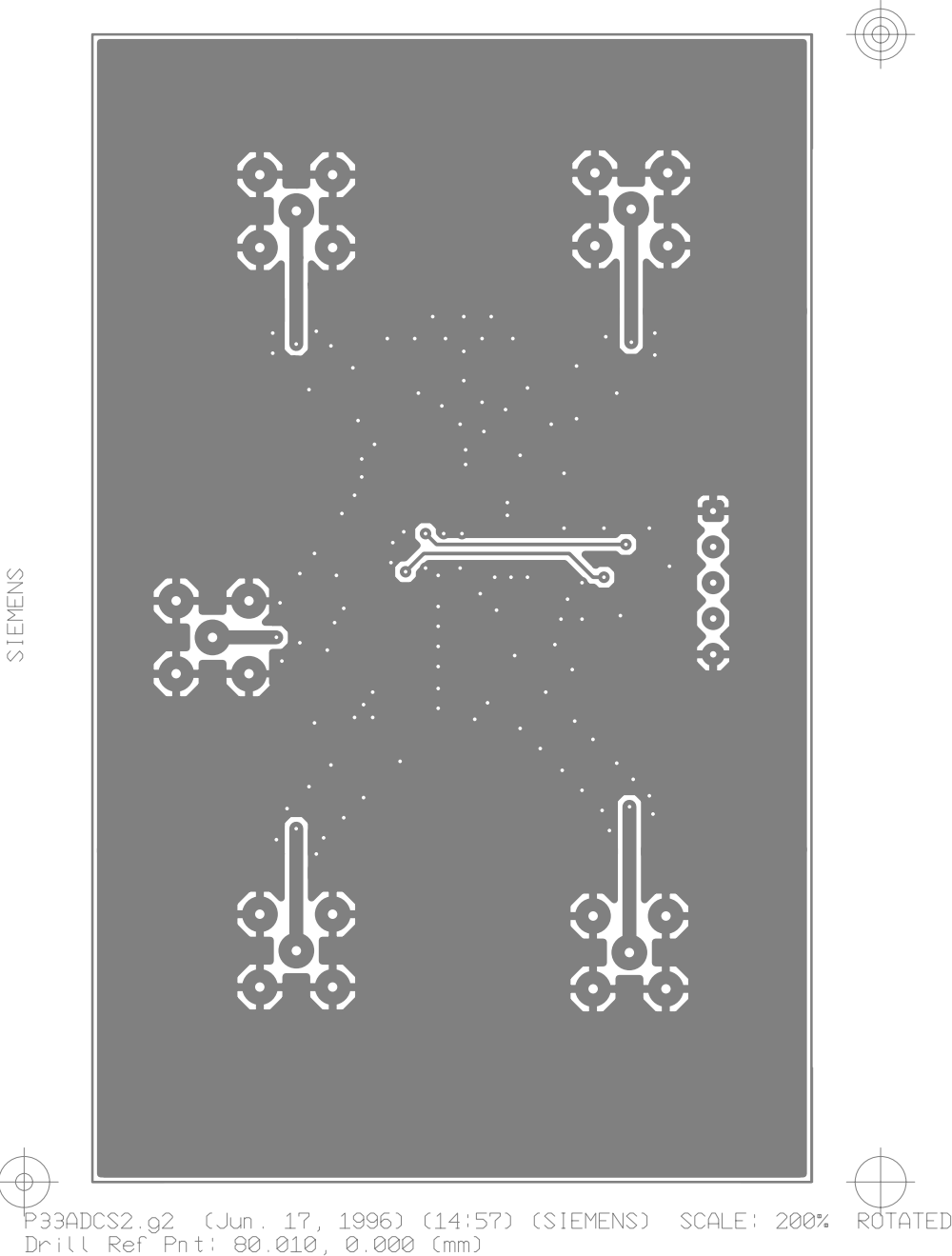


Figure 14: PCB bottom side

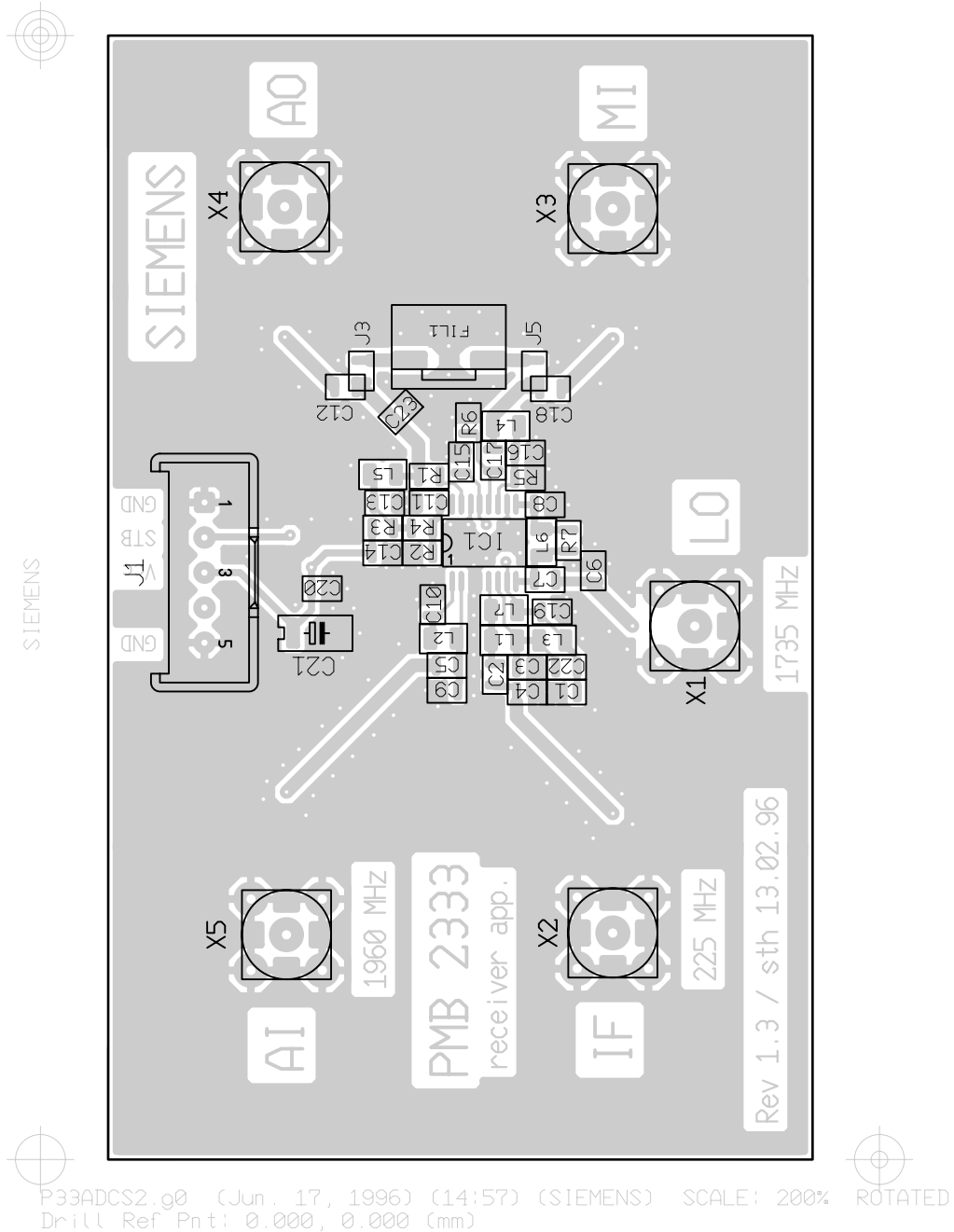


Figure 15: PCB component placement

List of Components

| Item | Quantity | Reference | Value | Part | |
|------|----------|--|------------------|--|-------------------------|
| 1 | 3 | C9, C17, C23 | 1.0 pF | SMD/0603 | |
| 2 | 1 | C6 | 1.5 pF | SMD/0603 | |
| 3 | 1 | C8 | 1.8 pF | SMD/0603 | |
| 4 | 1 | C16 | 2.2 pF | SMD/0603 | |
| 5 | 1 | C7 | 2.7 pF | SMD/0603 | |
| 6 | 1 | C22 | 4.7 pF | SMD/0603 | |
| 7 | 10 | C3, C5, C10, C11, C12, C13, C14, C15, C18, C19 | 10 pF | SMD/0603 | |
| 8 | 1 | C2 | 12 pF | SMD/0603 | |
| 9 | 1 | C4 | 33 pF | SMD/0603 | |
| 10 | 1 | C1 | 680 pF | SMD/0603 | |
| 11 | 1 | C20 | 1.0 nF | SMD/0603 | |
| 12 | 1 | C21 | 1 μ F | SMD/A | Tantalum |
| 13 | 1 | R1 | 0 Ω | SMD/0603 | |
| 14 | 3 | R3, R5, R6 | 180 Ω | SMD/0603 | |
| 15 | 2 | R4, R7 | 1.0 k Ω | SMD/0603 | |
| 16 | 1 | L6 | 3.3 nH | SMD/0805 | Murata LQP21A or LQP11A |
| 17 | 1 | L4 | 3.9 nH | SMD/0805 | Murata LQP21A or LQP11A |
| 18 | 1 | L5 | 6.8 nH | SMD/0805 | Murata LQP21A or LQP11A |
| 19 | 1 | L2 | 22 nH | SMD/0805 | Coilcraft 0805 |
| 20 | 1 | L1 | 82 nH | SMD/0805 | Coilcraft 0805 |
| 21 | 1 | L3 | 100 nH | SMD/0805 | Coilcraft 0805 |
| 22 | 1 | IC1 | PMB 2333 | Siemens | |
| 23 | 1 | J1 | connector | Stocko MKS 1655-6-0-505 | |
| 24 | 5 | X1, X2, X3, X4, X5 | SMA connector | Suhner 82 SMA 50-0-41 or Rosenberger 32 K 141-400A2 | |

3.2 Upconversion Application

3.2.1 Shortform Data

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 2.7\text{ V}$

Mixer input signal $f_{IF} = 190\text{ MHz}$, $P_{IF} = -30\text{ dBm}$

LO signal $f_{LO} = 1717\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

Mixer output and driver amplifier input signal $f_{RF} = 1907\text{ MHz}$

| Parameter | Symbol | Limit values | | | Unit | Remarks |
|---------------------------------|---------------|--------------|------|------|------|-------------|
| | | min. | typ. | max. | | |
| Mixer section | | | | | | |
| Mixer current | I_{Mixer} | | 9.5 | | mA | |
| Conversion gain | G_C | | 6 | | dB | |
| 3rd order input intercept point | IICP3 | | +3.5 | | dBm | |
| 1dB-compression point | P_{1dB} | | -7 | | dBm | |
| Port matching | | | | | | |
| RF return loss | $ S_{11,RF} $ | | 10 | | dB | |
| LO return loss | $ S_{11,LO} $ | | 18 | | dB | |
| IF return loss | $ S_{11,IF} $ | | 20 | | dB | |
| Isolations | | | | | | |
| RF to LO input | A_{RF-LO} | | 38 | | dB | |
| LO to RF output | A_{LO-RF} | | 30 | | dB | |
| Driver section | | | | | | |
| Driver current | I_{AO} | | 11.0 | | mA | 17mA @ 3.3V |
| Gain | G | | 13.1 | | dB | |
| 3rd order input intercept point | IICP3 | | +4.5 | | dBm | |
| 1dB-compression point | P_{1dB} | | -6 | | dBm | |

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 2.7\text{ V}$

Mixer input signal $f_{IF} = 190\text{ MHz}$, $P_{IF} = -30\text{ dBm}$

LO signal $f_{LO} = 1717\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

Mixer output and driver amplifier input signal $f_{RF} = 1907\text{ MHz}$

| Parameter | Symbol | Limit values | | | Unit | Remarks |
|----------------|---------------|--------------|------|------|------|---------|
| | | min. | typ. | max. | | |
| AI return loss | $ S_{11,AI} $ | | 14 | | dB | |
| AO return loss | $ S_{11,AO} $ | | 12 | | dB | |

3.2.2 Measurement results

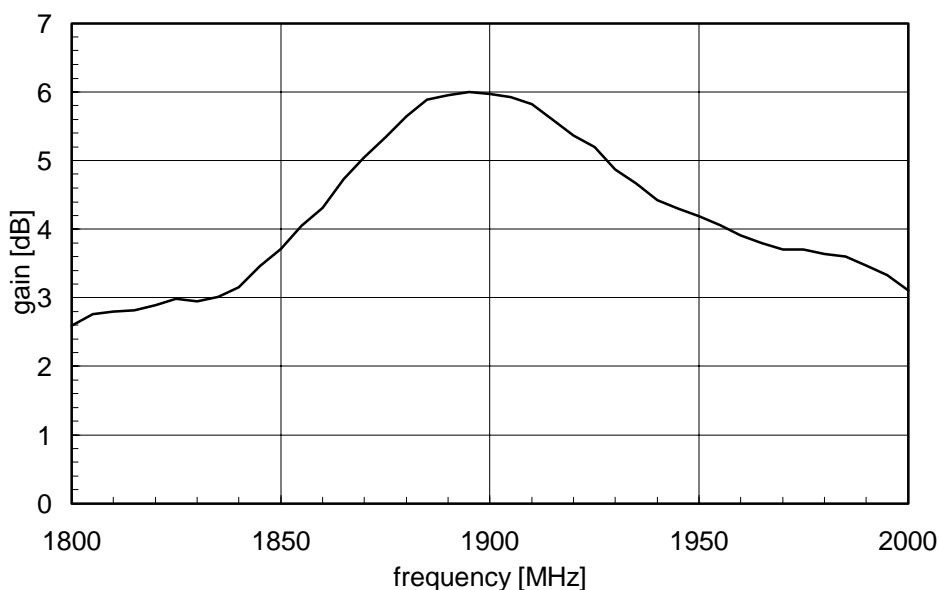


Figure 16: Mixer conversion gain versus frequency

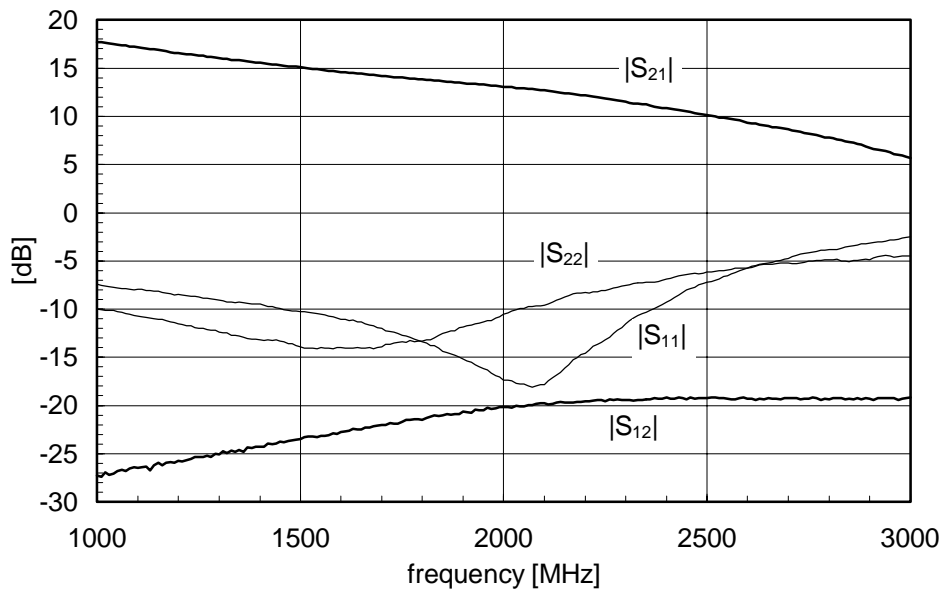


Figure 17: Driver amplifier gain, reverse isolation, return loss versus frequency

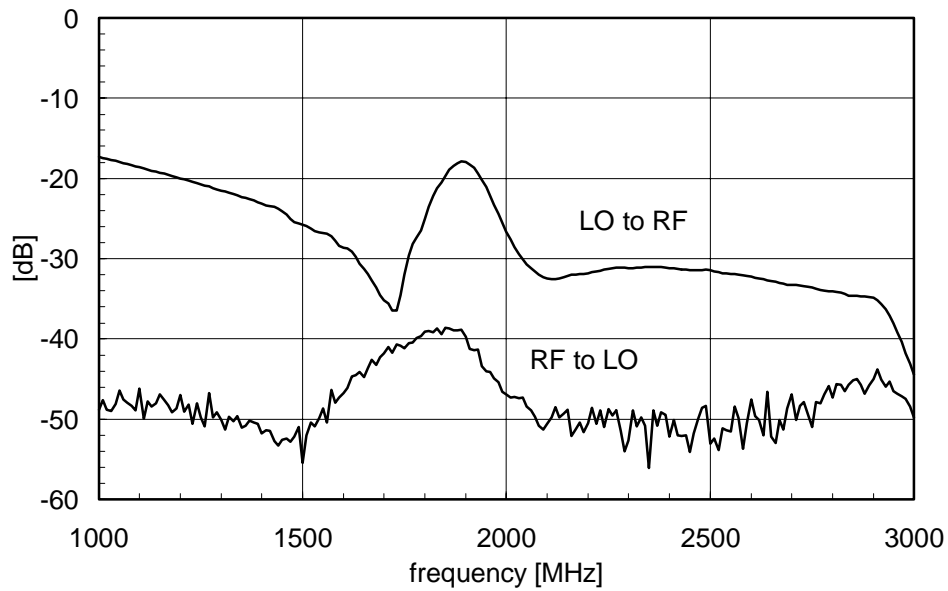


Figure 18: Mixer isolations versus frequency

3.2.3 Circuit diagram and PCB layout

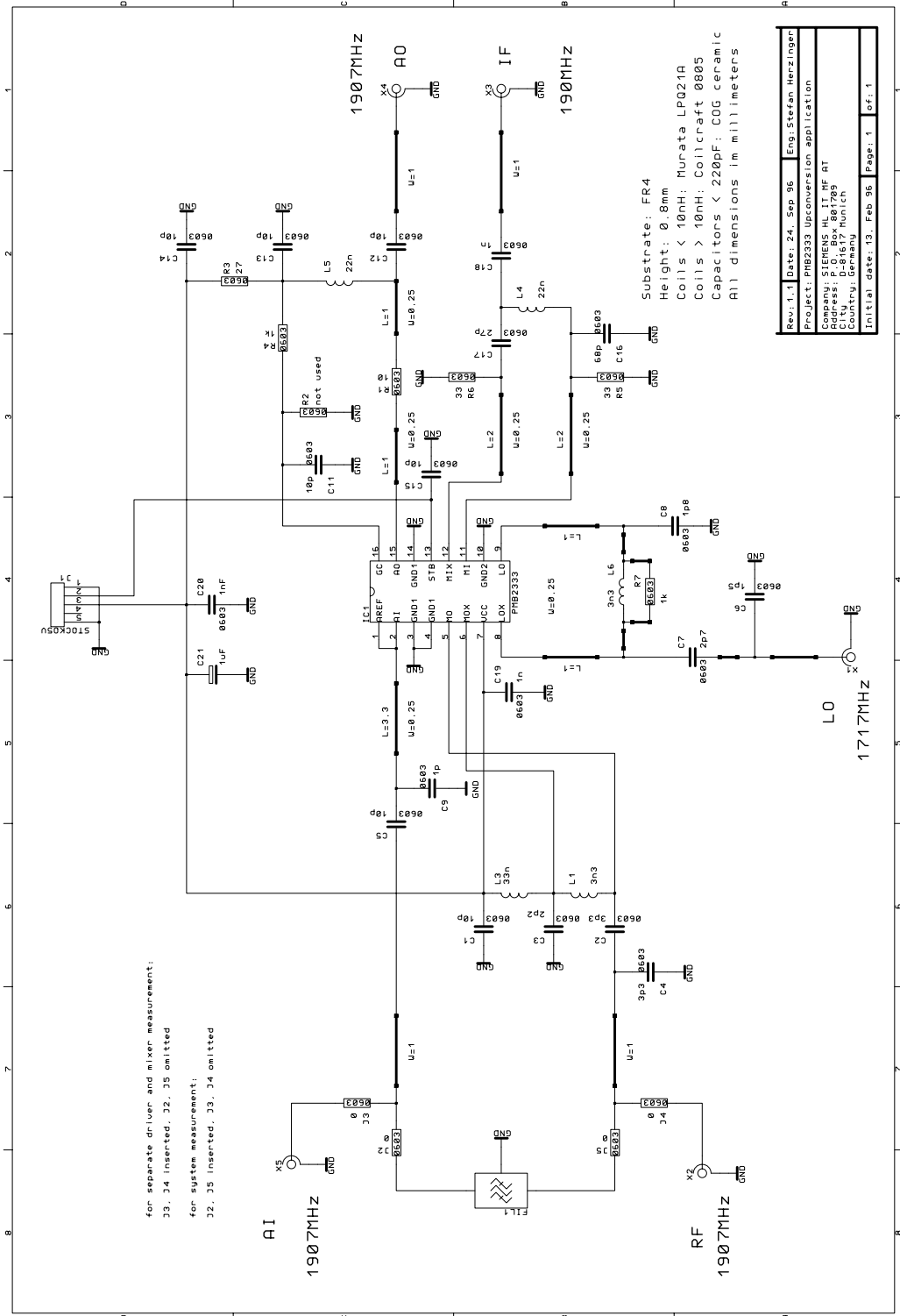


Figure 19: Circuit diagram

PCB dimensions: 80 x 50 mm

Substrate material: FR4

Substrate height: 0.8 mm

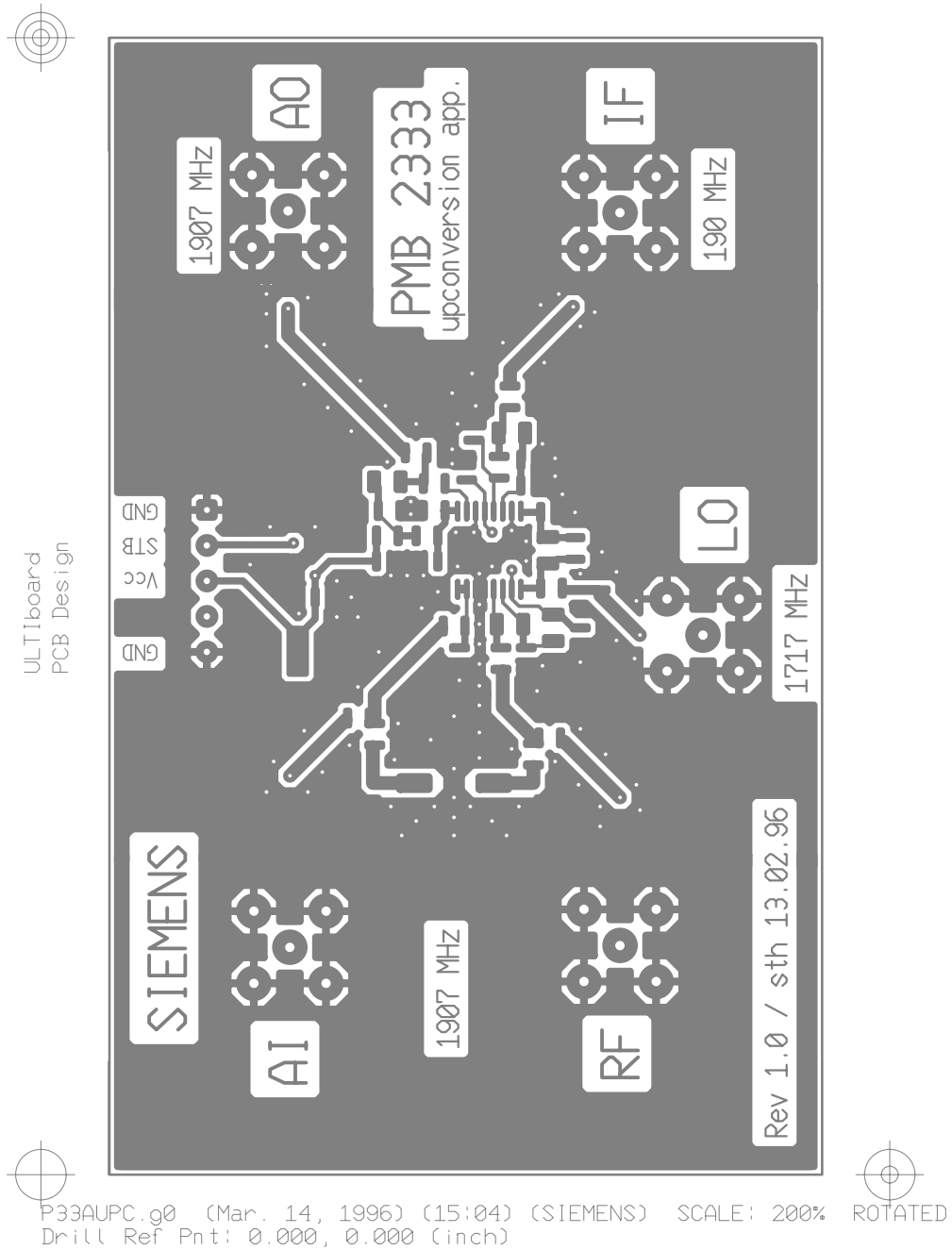


Figure 20: PCB top side

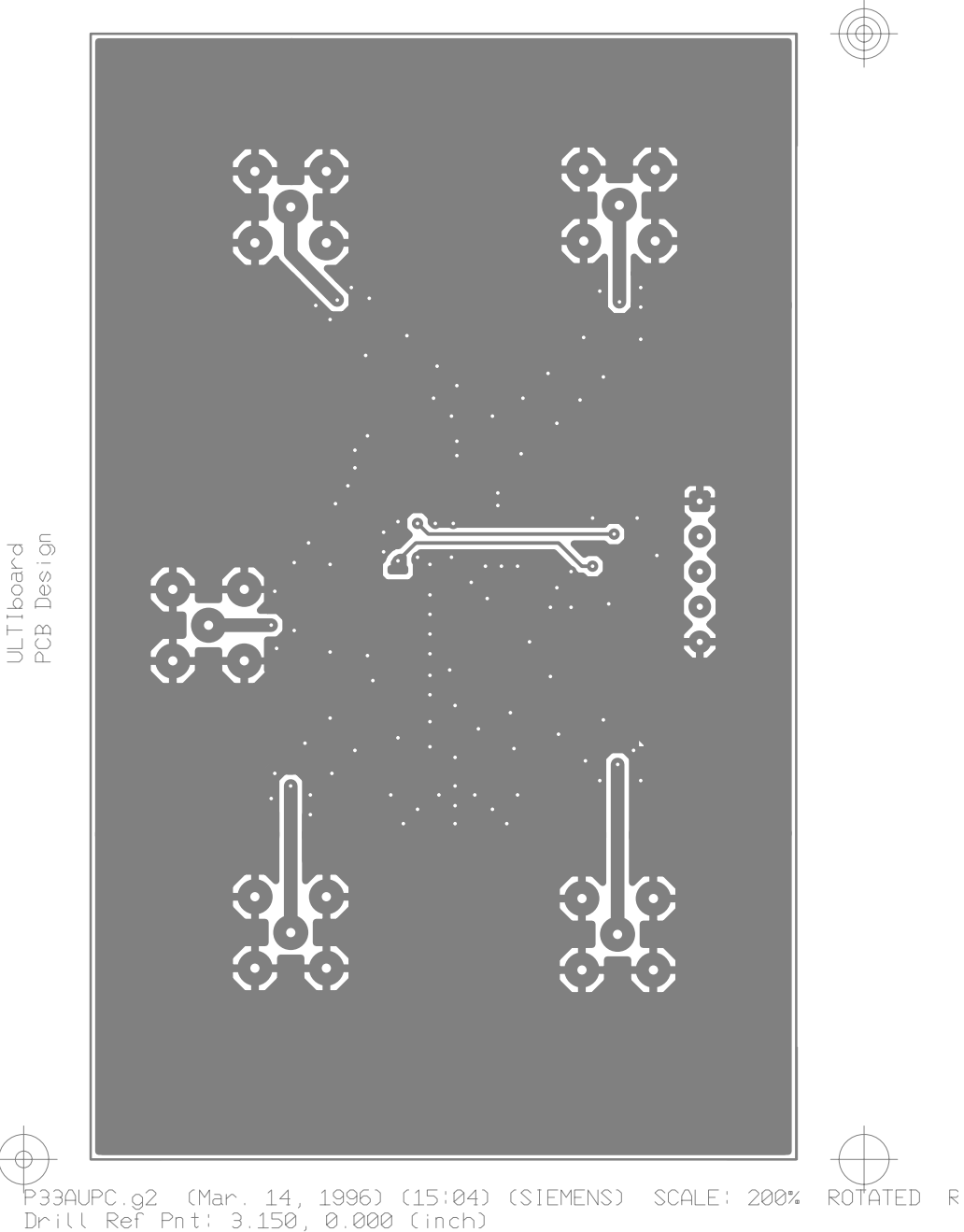


Figure 21: PCB bottom side

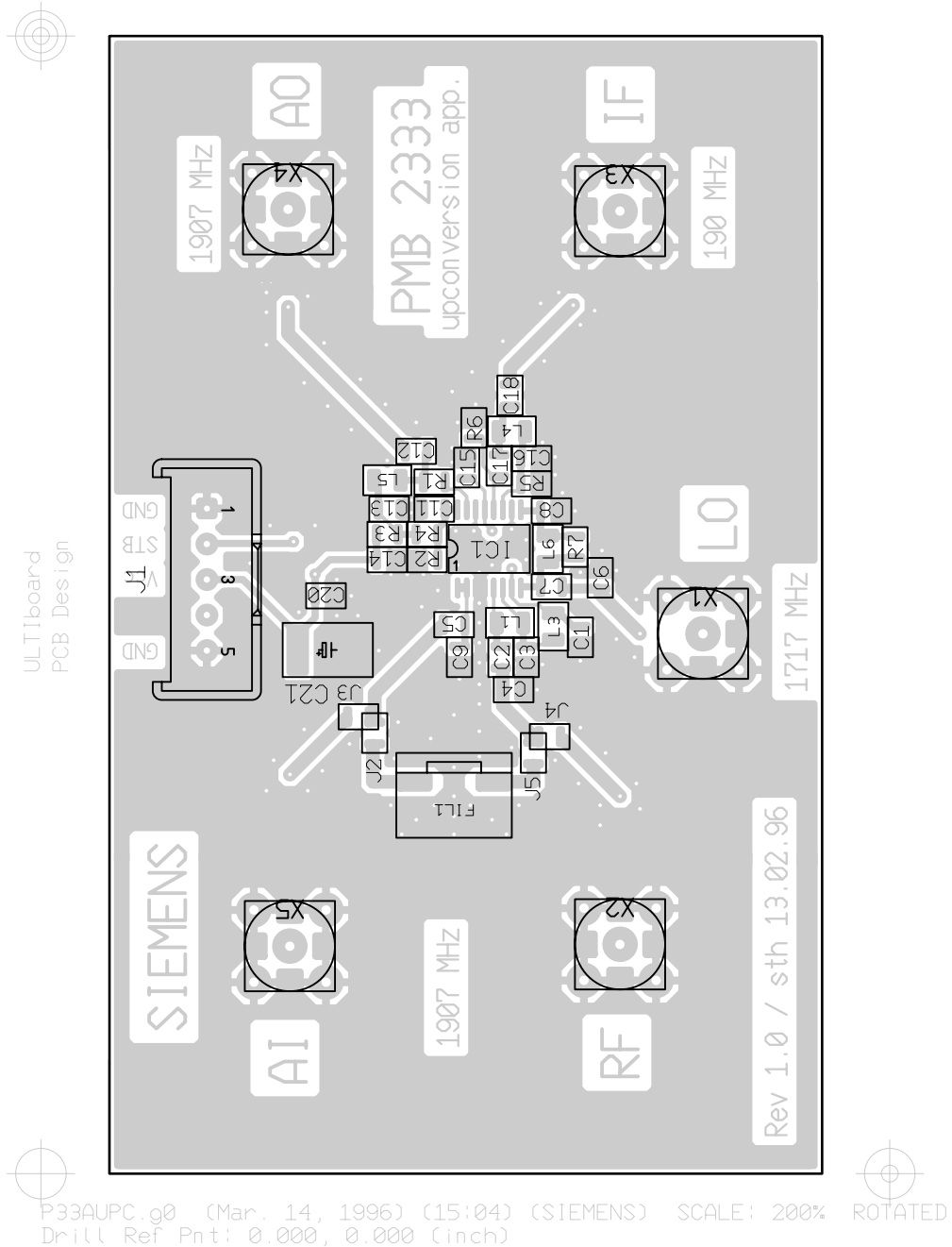


Figure 22: Component placement top side

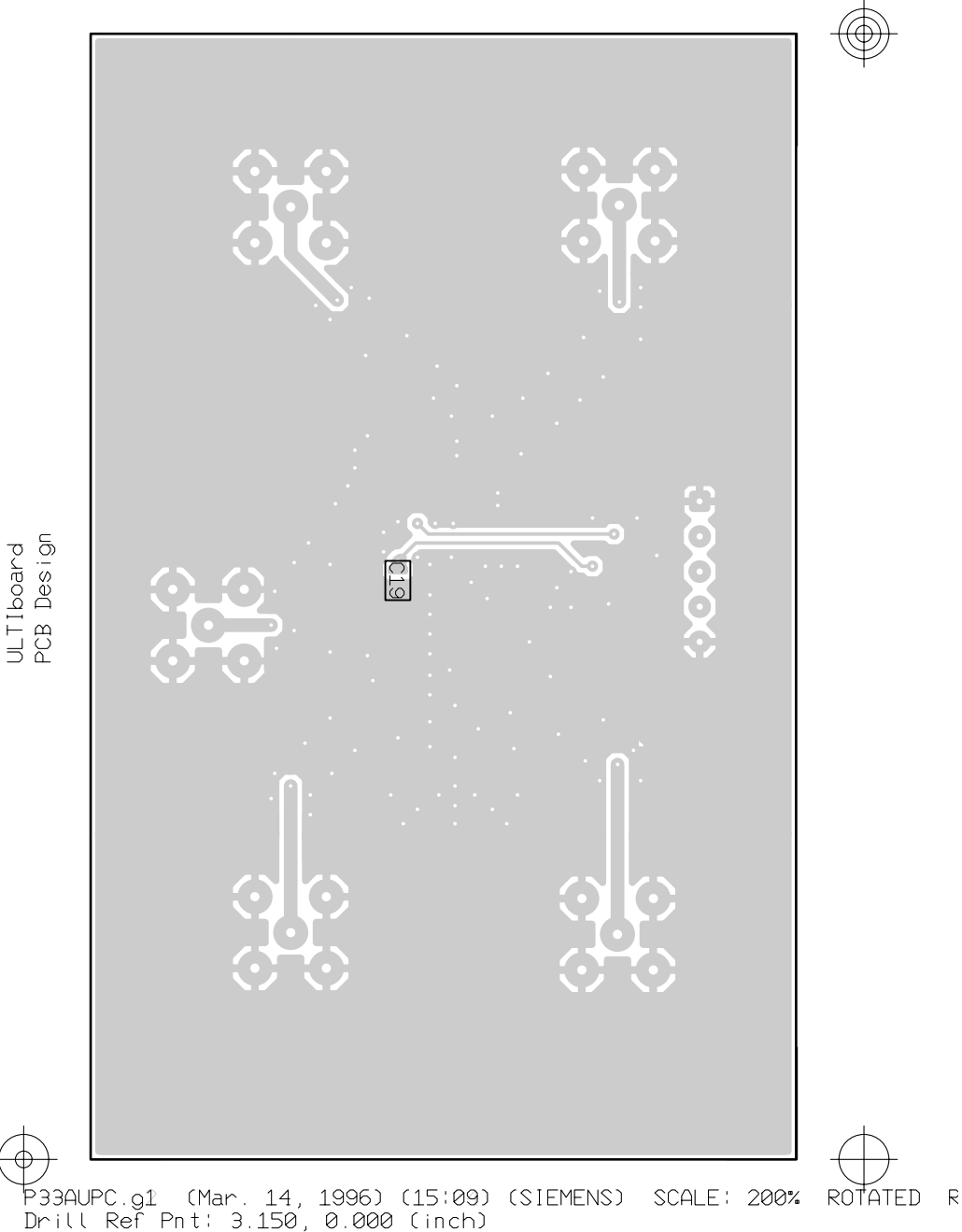


Figure 23: Component placement bottom side

Upconversion Application - List of Components

| Item | Quantity | Reference | Value | Part | |
|------|----------|--|------------------|--|----------------|
| 1 | 1 | C9 | 1.0 pF | SMD/0603 | |
| 2 | 1 | C6 | 1.5 pF | SMD/0603 | |
| 3 | 1 | C8 | 1.8 pF | SMD/0603 | |
| 4 | 1 | C3 | 2.2 pF | SMD/0603 | |
| 5 | 1 | C7 | 2.7 pF | SMD/0603 | |
| 6 | 2 | C2, C4 | 3.3 pF | SMD/0603 | |
| 7 | 7 | C1, C5, C11, C12, C13, C14, C15 | 10 pF | SMD/0603 | |
| 8 | 1 | C17 | 27 pF | SMD/0603 | |
| 9 | 1 | C16 | 68 pF | SMD/0603 | |
| 10 | 3 | C18, C19, C20 | 1.0 nF | SMD/0603 | |
| 11 | 1 | C21 | 1.0 μ F | SMD/A | Tantalum |
| 12 | 2 | J3, J4 | 0 Ω | SMD/0603 | |
| 13 | 1 | R1 | 10 Ω | SMD/0603 | |
| 14 | 1 | R3 | 27 Ω | SMD/0603 | |
| 15 | 2 | R5, R6 | 33 Ω | SMD/0603 | |
| 16 | 2 | R4, R7 | 1 k Ω | SMD/0603 | |
| 17 | 2 | L1, L6 | 3.3 nH | SMD/0603 | Murata LQP11A |
| 18 | 2 | L4, L5 | 22 nH | SMD/0805 | Coilcraft 0805 |
| 19 | 1 | L3 | 33 nH | SMD/0805 | Coilcraft 0805 |
| 20 | 1 | IC1 | PMB 2333 | Siemens | |
| 21 | 1 | J1 | connector | Stocko MKS 1655-6-0-505 | |
| 22 | 5 | X1, X2, X3, X4, X5 | SMA connector | Suhner 82 SMA 50-0-41 or Rosenberger 32 K 141-400A2 | |

3.3 Receiver/SAW Application

3.3.1 Shortform Data

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 3.6\text{ V}$

LNA and Mixer input signal $f_{RF} = 890\text{ MHz}$, $P_{RF} = -40\text{ dBm}$

LO signal $f_{LO} = 972\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

IF output $f_{IF} = 82\text{ MHz}$

All measurements refer to SMA connectors without consideration of PCB losses

The figures in the mixer section are calculated from a measurement SAW filter + mixer assuming an insertion loss of 3.1dB for the filter at measurement frequency (890MHz).

| Parameter | Symbol | typ. Value | Unit | Remarks |
|---------------------------------|----------------|------------|------|------------------|
| Mixer section | | | | |
| Mixer current | I_{Mixer} | 10 | mA | $I_{MO}+I_{MOX}$ |
| Conversion gain | G_C | 4.5 | dB | |
| Noise Figure (SSB) | NF_{SSB} | 12 | dB | |
| 3rd order input intercept point | $IICP3$ | +6.5 | dBm | |
| input 1dB-compression point | P_{1dB} | -3 | dBm | |
| LNA section | | | | |
| LNA current | I_{LNA} | 17 | mA | |
| Gain | G_{LNA} | 20 | dB | |
| Noise Figure | NF | 1.75 | dB | |
| 3rd order input intercept point | $IICP3$ | 0 | dBm | |
| input 1dB-compression point | P_{1dB} | -11 | dBm | |
| AI return loss | $ S_{11, AI} $ | 14 | dB | |
| AO return loss | $ S_{11, AO} $ | 15 | dB | |
| | | | | |
| | | | | |
| | | | | |

Measurement conditions

Ambient temperature $T_A = 25\text{ °C}$

Supply voltage $V_S = 3.6\text{ V}$

LNA and Mixer input signal $f_{RF} = 890\text{ MHz}$, $P_{RF} = -40\text{ dBm}$

LO signal $f_{LO} = 972\text{ MHz}$, $P_{LO} = -6\text{ dBm}$

IF output $f_{IF} = 82\text{ MHz}$

All measurements refer to SMA connectors without consideration of PCB losses

The figures in the mixer section are calculated from a measurement SAW filter + mixer assuming an insertion loss of 3.1dB for the filter at measurement frequency (890MHz).

| Parameter | Symbol | typ. Value | Unit | Remarks |
|--|-----------|------------|------|---------|
| Cascade figures | | | | |
| LNA, resistive 3.1 dB pad (see schematic), SAW filter, mixer | | | | |
| Cascade Gain | G | 18 | dB | |
| Cascade Noise Figure (SSB) | NF | 3.35 | dB | |
| 3rd order input intercept point | IICP3 | -8 | dBm | |
| input 1dB-compression point | P_{1dB} | -17 | dBm | |

3.3.2 System calculations

Due to the use of a SAW filter for mixer input matching and balancing an isolated measurement of the mixer figures is not possible. The following system calculations have valid entries only for the Gain, IP3, and Noise Figure. The input values are either from extra measurements (e.g. SAW filter insertion loss) or adjusted to give a cascade figure that can be measured (e.g. filter+mixer IP3). All non-fat typeface figures have no meaning.

LNA measurement

| Hewlett-Packard | NoiseCalc | | AppCAD | |
|------------------------------|--------------|--------------|------------------------------|--------------|
| | +----+ | +----+ | | |
| | - 1 +--- | 2 +- | | |
| | +----+ | +----+ | | |
| Noise Figure (dB) | 0.10 | 1.65 | | |
| Gain (dB) | -0.10 | 20.00 | | |
| IP3 (dBm) | 100.00 | 20.00 | | |
| System Temp. (°C) | 25.0 | | Reference Temperature (°C) | 25.0 |
| Input Power (dBm) | -30.0 | | Noise Bandwidth (MHz) | 1.00000 |
| Pout (dBm) | -30.1 | -10.1 | | |
| Cascade NF (dB) | 1.75 | | Cascade Gain (dB) | 19.90 |
| Noise Temperature (°K) | 143.9 | | Input Intercept Point (dBm) | 0.1 |
| Signal-to-Noise Ratio (dB) | 82.2 | | Output Intercept Point (dBm) | 20.0 |
| Spur Free Dynamic Range (dB) | 74.9 | | IM3 Output Level (dBm) | -70.3 |
| Nominal Detectable Sig (dBm) | -112.2 | | | |

- 1) PCB loss
- 2) PMB 2333 LNA

SAW filter and Mixer

| Hewlett-Packard | NoiseCalc | | AppCAD | |
|------------------------------|---------------|--------------|------------------------------|-------------|
| | +----+ | +----+ | | |
| | - 1 +--- | 2 +- | | |
| | +----+ | +----+ | | |
| Noise Figure (dB) | 3.10 | 12.00 | | |
| Gain (dB) | -3.10 | 4.50 | | |
| IP3 (dBm) | 100.00 | 11.00 | | |
| System Temp. (°C) | 25.0 | | Reference Temperature (°C) | 25.0 |
| Input Power (dBm) | -30.0 | | Noise Bandwidth (MHz) | 1.00000 |
| Pout (dBm) | -33.1 | -28.6 | | |
| Cascade NF (dB) | 15.10 | | Cascade Gain (dB) | 1.40 |
| Noise Temperature (°K) | 9094.2 | | Input Intercept Point (dBm) | 9.6 |
| Signal-to-Noise Ratio (dB) | 68.9 | | Output Intercept Point (dBm) | 11.0 |
| Spur Free Dynamic Range (dB) | 72.3 | | IM3 Output Level (dBm) | -107.8 |
| Nominal Detectable Sig (dBm) | -98.9 | | | |

- 1) SAW filter S+M B4672
- 2) PMB 2333 Mixer

Overall figures

| Hewlett-Packard | NoiseCalc | | | | | AppCAD | |
|------------------------------|-----------|--------|------------------------------|----------------------------|---------|--------|--|
| | +----+ | +----+ | +----+ | +----+ | +----+ | | |
| | - 1 +--- | 2 +--- | 3 +--- | 4 +--- | 5 +- | | |
| | +----+ | +----+ | +----+ | +----+ | +----+ | | |
| Noise Figure (dB) | 0.10 | 1.65 | 3.10 | 3.10 | 12.00 | | |
| Gain (dB) | -0.10 | 20.00 | -3.10 | -3.10 | 4.50 | | |
| IP3 (dBm) | 200.00 | 20.00 | 200.00 | 40.00 | 11.00 | | |
| System Temp. (°C) | 25.0 | | | Reference Temperature (°C) | 25.0 | | |
| Input Power (dBm) | -30.0 | | | Noise Bandwidth (MHz) | 1.00000 | | |
| Pout (dBm) | -30.1 | -10.1 | -13.2 | -16.3 | -11.8 | | |
| Cascade NF (dB) | 3.35 | | Cascade Gain (dB) | 18.20 | | | |
| Noise Temperature (°K) | 337.0 | | Input Intercept Point (dBm) | -7.9 | | | |
| Signal-to-Noise Ratio (dB) | 80.7 | | Output Intercept Point (dBm) | 10.3 | | | |
| Spur Free Dynamic Range (dB) | 68.5 | | IM3 Output Level (dBm) | -55.9 | | | |
| Nominal Detectable Sig (dBm) | -110.6 | | | | | | |

- 1) PCB loss
- 2) PMB 2333 LNA
- 3) Resistive pad, 3.1dB attenuation
- 4) SAW filter S+M B4672
- 5) PMB 2333 Mixer

3.3.3 Measurement results

Cascaded figure measurements

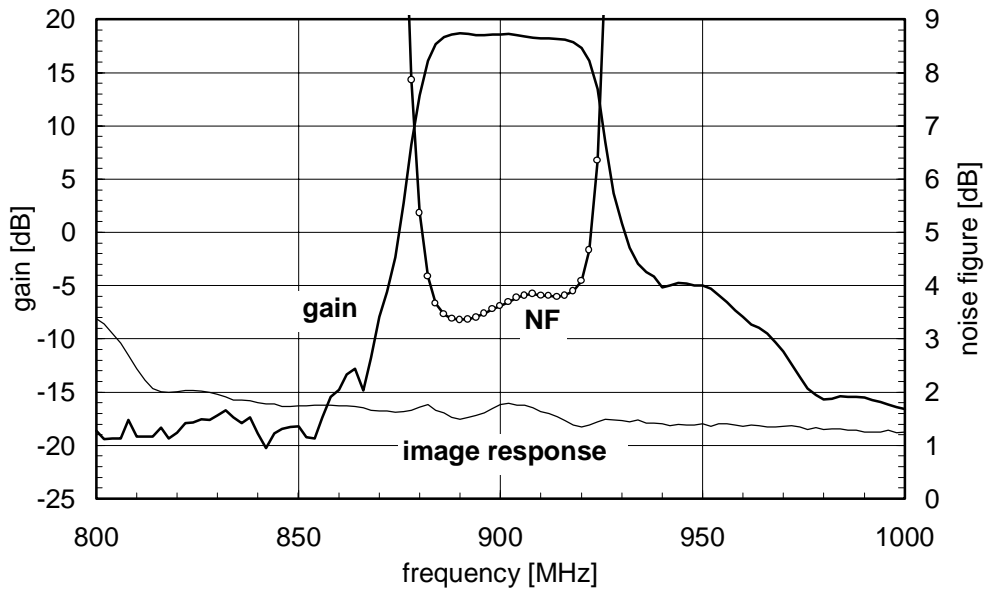


Figure 24: Cascade gain, SSB noise figure versus frequency

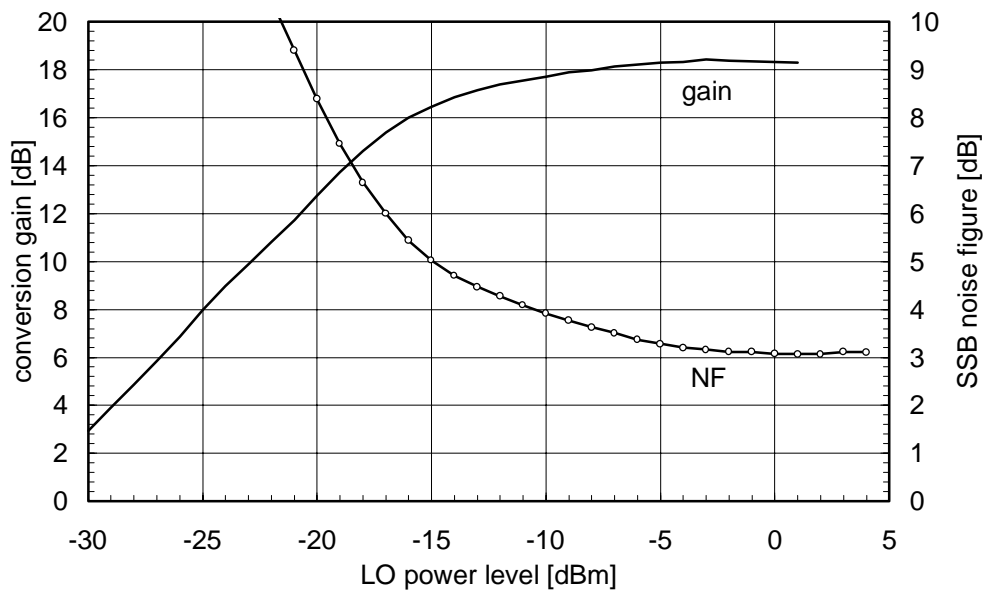


Figure 25: Cascade gain, SSB noise figure versus LO power

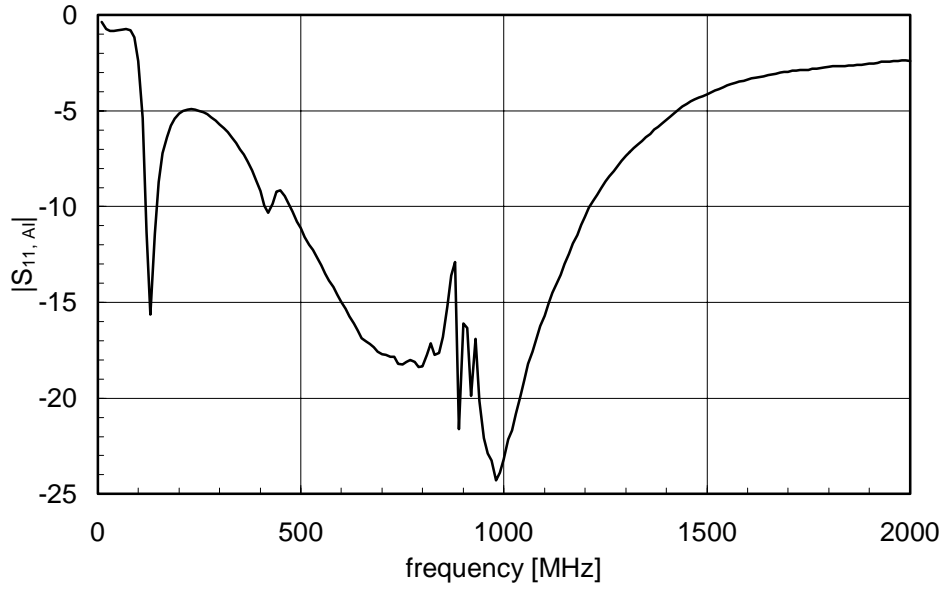


Figure 26: AI input return loss versus frequency

3.3.4 Circuit diagram and PCB layout

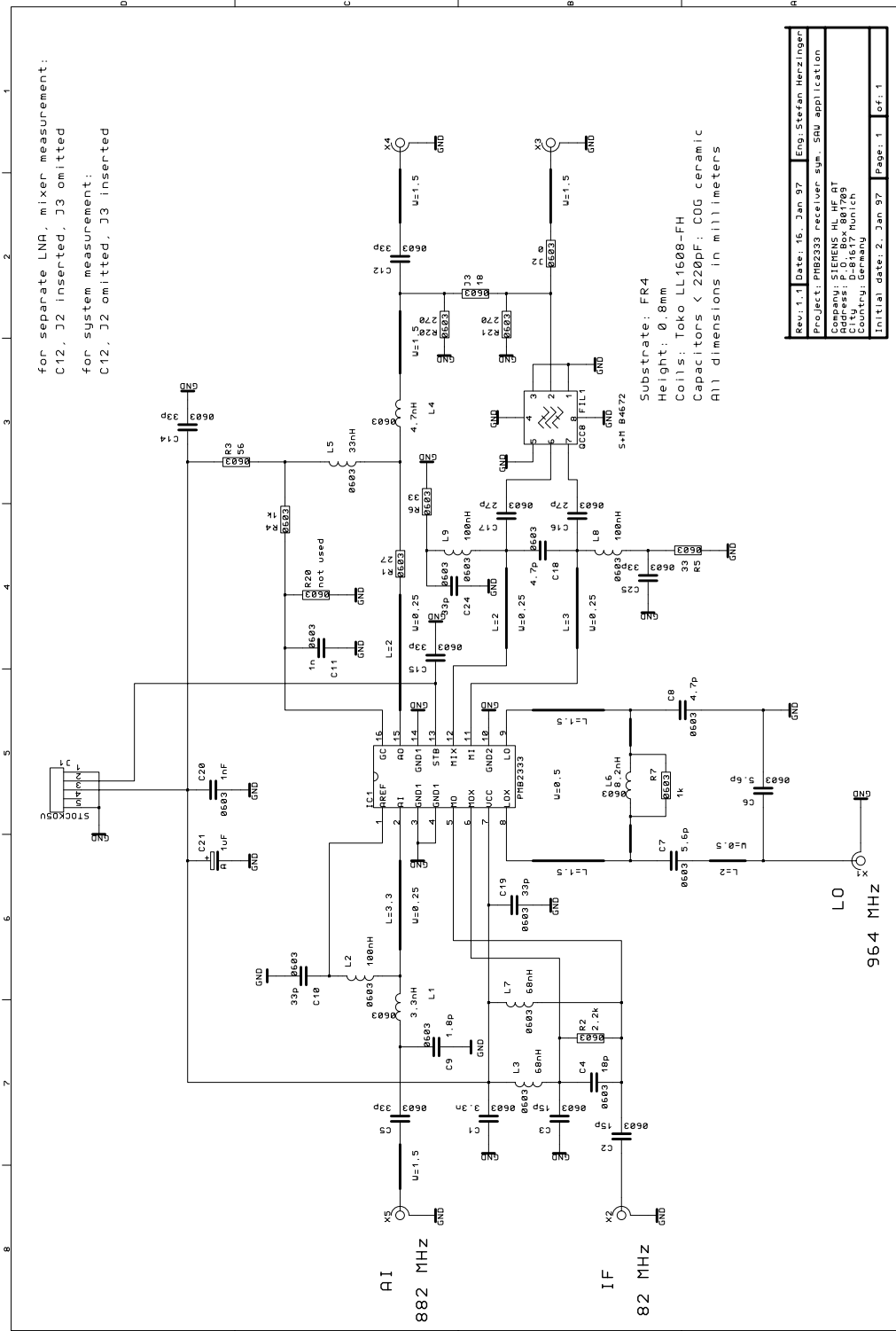


Figure 27: Circuit diagram

PCB dimensions: 80 x 50 mm

Substrate material: FR4

Substrate height: 0.8 mm

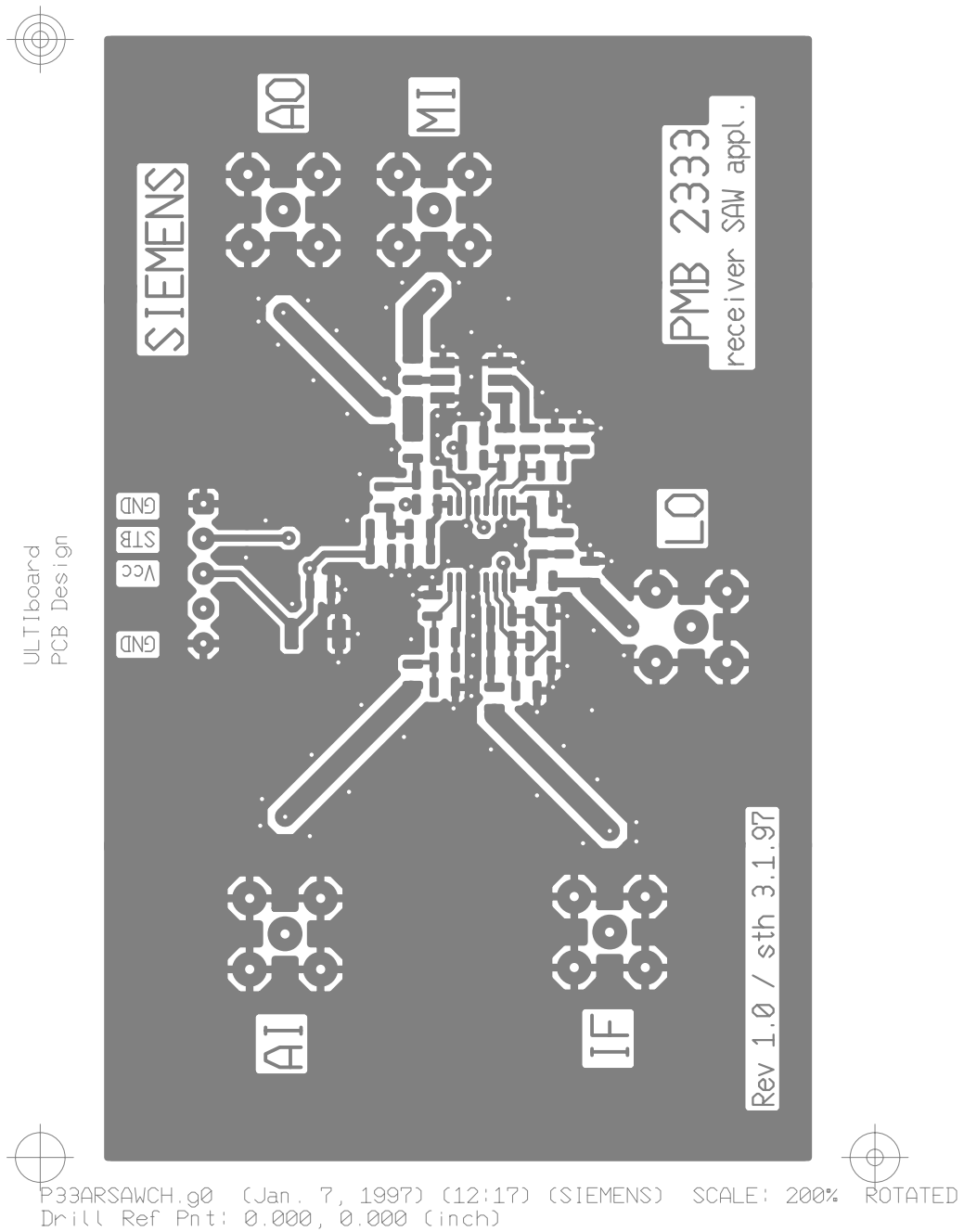


Figure 28: PCB top side

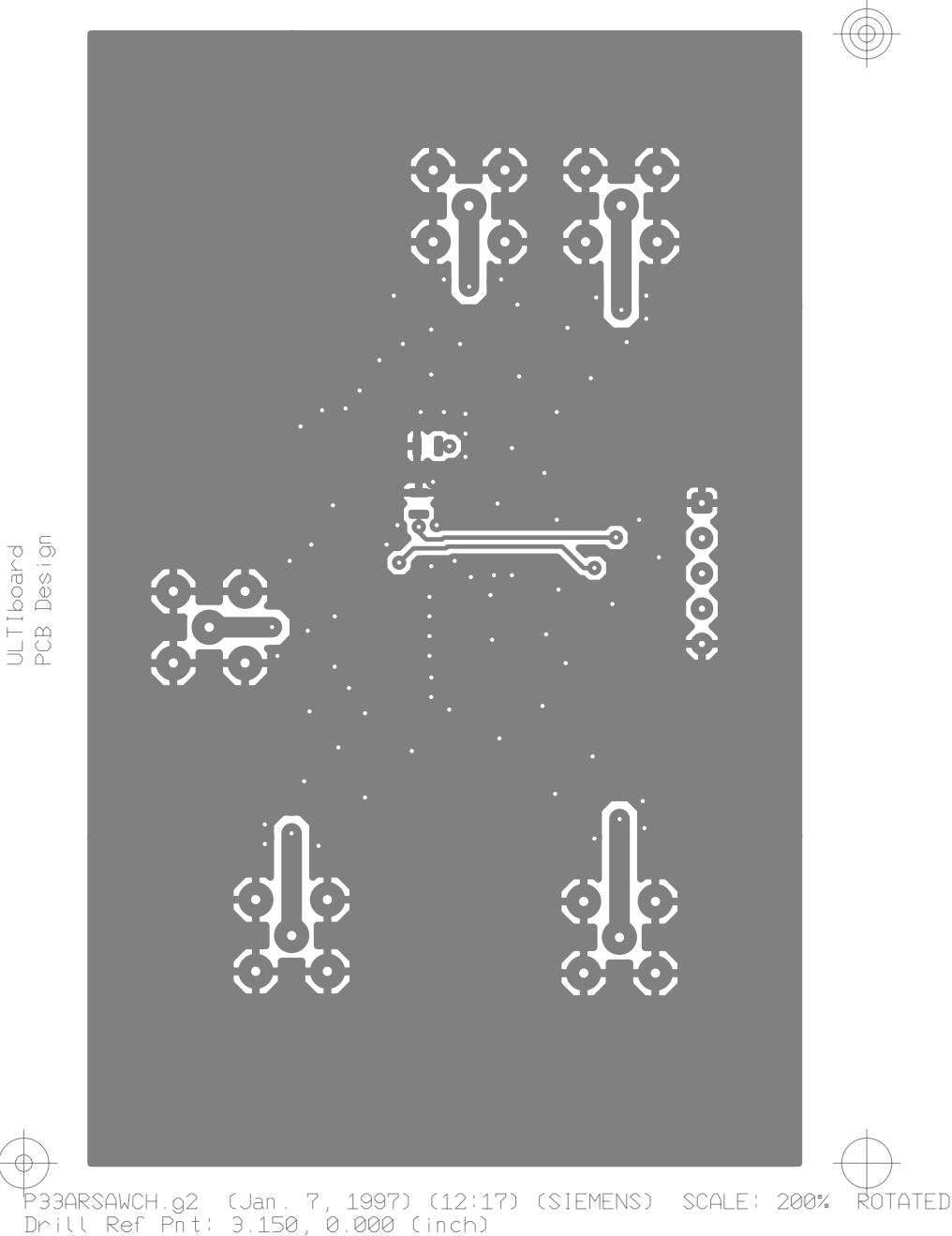


Figure 29: PCB bottom side

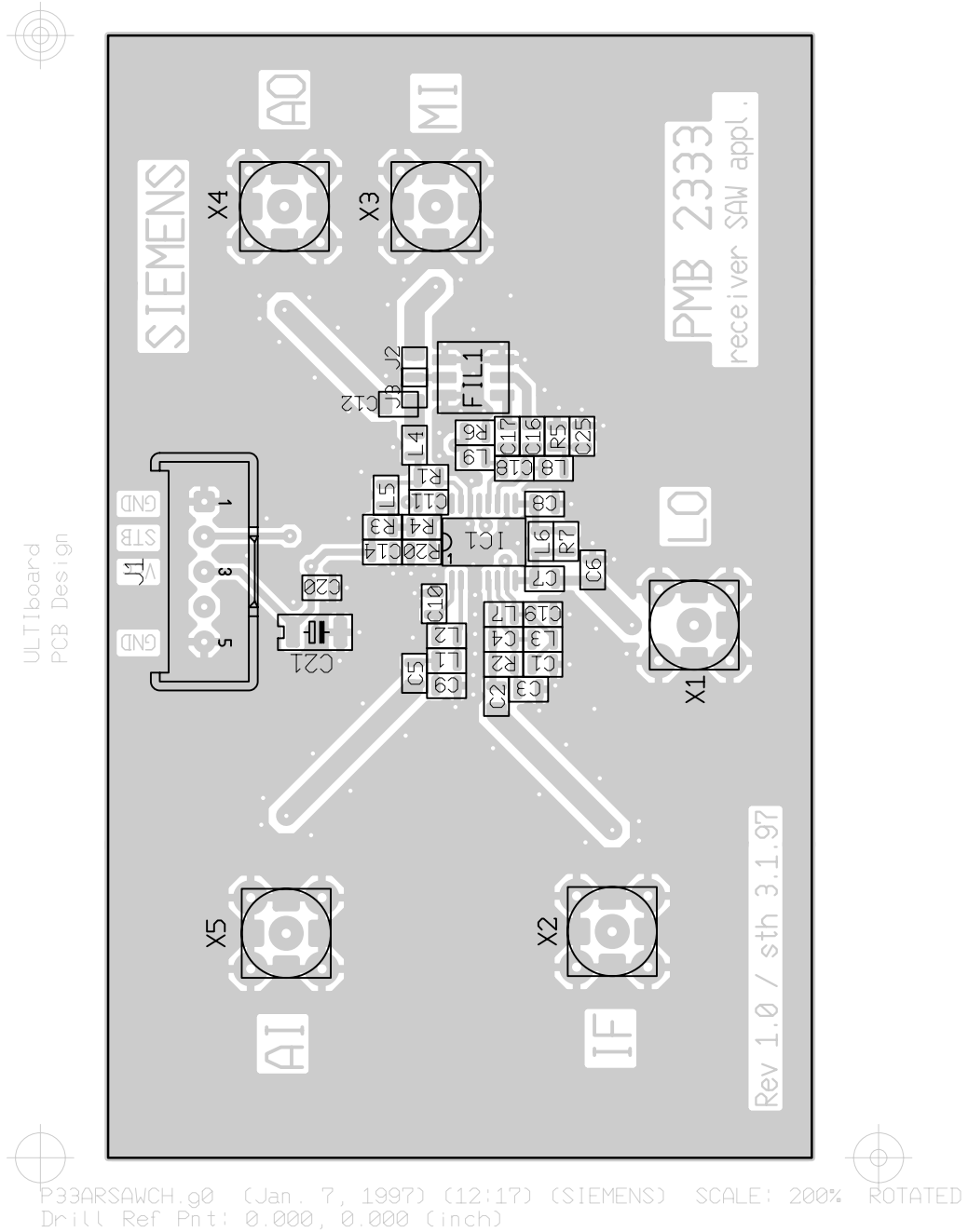


Figure 30: PCB component placement top

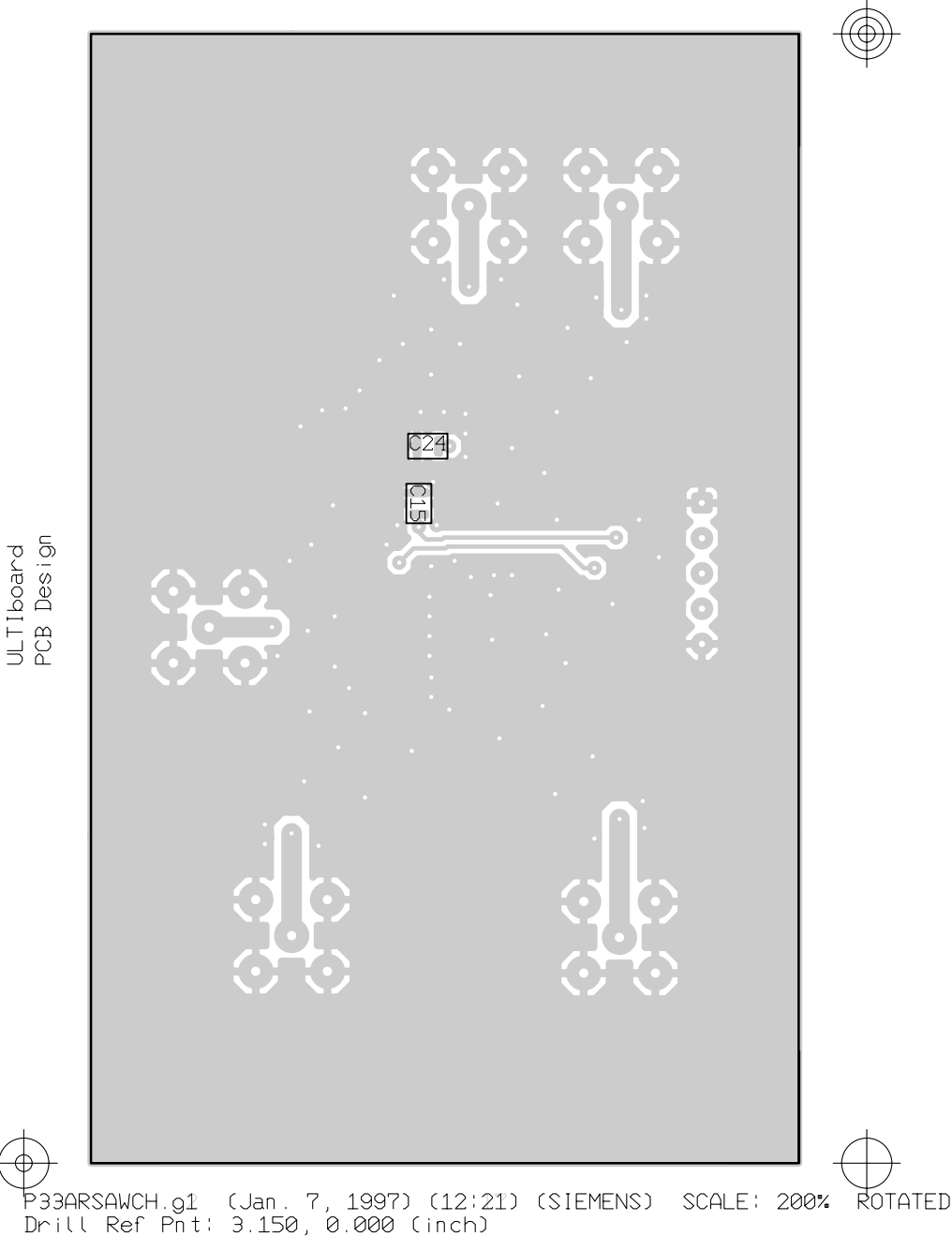


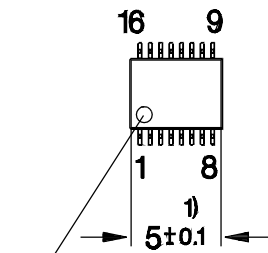
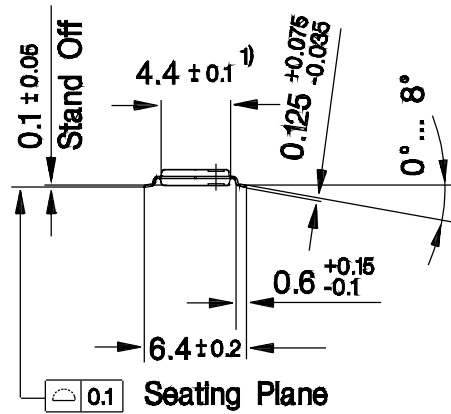
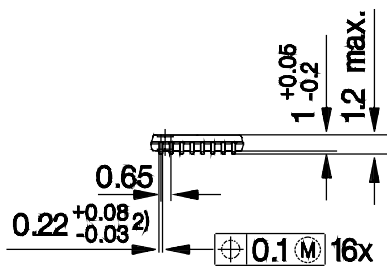
Figure 31: PCB component placement bottom

Receiver/SAW Application - List of Components

| Item | Quantity | Reference | Value | Part | |
|------|----------|--|------------------|--|----------------|
| 1 | 1 | C9 | 1.8 pF | SMD/0603 | |
| 2 | 2 | C8, C18 | 4.7 pF | SMD/0603 | |
| 3 | 2 | C6, C7 | 5.6 pF | SMD/0603 | |
| 4 | 2 | C2, C3 | 15 pF | SMD/0603 | |
| 5 | 1 | C4 | 18 pF | SMD/0603 | |
| 6 | 2 | C16, C17 | 27 pF | SMD/0603 | |
| 7 | 8 | C5, C10, C12, C14, C15, C19, C24, C25 | 33 pF | SMD/0603 | |
| 8 | 2 | C11, C20 | 1 nF | SMD/0603 | |
| 9 | 1 | C1 | 3.3 nF | SMD/0603 | |
| 10 | 1 | C21 | 1 μ F | SMD/A | Tantalum |
| 11 | 1 | J2 | 0 Ω | SMD/0603 | |
| 12 | 1 | R1 | 27 Ω | SMD/0603 | |
| 13 | 2 | R5, R6 | 33 Ω | SMD/0603 | |
| 14 | 1 | R3 | 56 Ω | SMD/0603 | |
| 15 | 2 | R4, R7 | 1.0 k Ω | SMD/0603 | |
| 16 | 1 | L1 | 3.3 nH | SMD/0603 | Toko LL1608-FH |
| 17 | 1 | L4 | 4.7 nH | SMD/0603 | Toko LL1608-FH |
| 18 | 1 | L6 | 8.2 nH | SMD/0603 | Toko LL1608-FH |
| 19 | 1 | L5 | 33 nH | SMD/0603 | Toko LL1608-FH |
| 20 | 2 | L3, L7 | 68 nH | SMD/0805 | Toko LL1608-FH |
| 21 | 3 | L2, L8, L9 | 100 nH | SMD/0805 | Toko LL1608-FH |
| 22 | 1 | IC1 | PMB 2333 | Siemens | |
| 23 | 1 | FIL1 | B4672 | S+M | |
| 24 | 1 | J1 | connector | Stocko MKS 1655-6-0-505 | |
| 25 | 5 | X1, X2, X3, X4, X5 | SMA connector | Suhner 82 SMA 50-0-41 or Rosenberger 32 K 141-400A2 | |

4 Package Outlines

P-TSSOP-16
(Plastic Package)



Index Marking

- 1) Does not include plastic or metal protrusion of 0.15 max. per side
- 2) Does not include dambar protrusion of 0.08 max. per side

Sorts of Packing

Package outlines for tubes, trays etc. are contained in our Data Book "Package Information".

SMD = Surface Mounted Device

Dimensions in mm